

Mobile Power Station

Final Report

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Table of Contents

1. Introduction Material.....	1
1.1 Executive Summary.....	1
1.2 Acknowledgements.....	1
1.3 Problem Statement	1
1.4 Potential solution	2
1.5 Operating Environment.....	2
1.6 Intended Users	2
1.7 Intended Uses.....	2
1.9 Market Survey.....	2
1.10 Intellectual Property and Legal Considerations.....	3
2. Approach	4
2.1 Concept Sketch	4
2.2 System Block Diagram	4
2.3 Functional Requirements	5
2.4 Non-functional Requirements	5
2.5 Assumptions and Limitations	5
2.6 Deliverables.....	6
2.7 Safety Considerations	6
3 Design, Implementation & Testing.....	7
3.1 Design.....	7
3.1.1 Power Inputs.....	7
3.1.2 MPPT.....	7
3.1.2.1 Li-ion Charging Cycle.....	9
3.1.3 Microprocessor	10
3.1.3.1 VCC and VSS.....	10
3.1.3.2 Voltage Senses	11
3.1.3.3 Low-side Current Sense	12
3.1.3.4 High-side Current Sense	13
3.1.3.5 UART Output	14
3.1.4 Charging Circuitry.....	14
3.1.5 Batteries	15
3.1.6 Buck Converter.....	15
3.2 Implementation	17
3.2.1 Initial Buck Converter	17

3.2.2 Configuring Microprocessor to Control Buck Converter.....	17
3.2.3 Printed Circuit Board.....	19
3.3 Testing	21
3.1 PCB Tests	21
3.2 Secondary Li-Ion Protection Testing	21
3.3 GUI Interface.....	22
3.4 Evaluation	23
4 Resource Requirements	24
4.1 Team Utilization.....	24
4.2 Material list and estimates	26
4.3 Human Resource Estimates	26
4.4 Project Schedule	27
4.5 Deliverables.....	27
5 Project Team Information.....	28
6. Closing Summary.....	29
References	30
Appendix I – Microprocessor Code	31
Appendix II – PCB Sensor Calibration Graphs	40
Appendix III – Schematic and PCB Layout	43
Appendix IV – User Manual	47

Table of Figures

Figure 1: MPS Concept Sketch	4
Figure 2: MPS Block Diagram	4
Figure 3: Current vs. Voltage Curve with Maximum Power Point [1].....	7
Figure 4: MPPT Perturbation and Observation Method [7]	8
Figure 5: Li-ion Charging Cycle.....	9
Figure 6: Microprocessor Schematic.....	10
Figure 7: Voltage Sense Circuit.....	11
Figure 8: Low-side Current Sense Circuit	12
Figure 9: High-side Current Sense Circuit	13
Figure 10: TTL/RS-232 Driver	14
Figure 11: Buck Converter Schematic	15
Figure 12: Large Transients observed at output of breadboard circuit	18
Figure 13: Prototype PCB.....	19
Figure 14: Populated PCB Rev. 1	20
Figure 15: Populated PCB Rev. 2.....	20
Figure 16: Software to Track and Monitor Cell Balancing	21
Figure 17: MPS GUI	22
Figure 18: Efficiency vs. PWM.....	23
Figure 19: Efficiency vs. Iout.....	23
Figure 20 - Human Resources Pie Chart.....	26
Figure 21: Semester 2 planned schedule.....	27

Table of Tables

Table 1: Electrical Parts Cost Per Unit Estimate	26
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1. Introduction Material

1.1 Executive Summary

With mobile technology surrounding every aspect of our lives, the need for mobile power—especially in remote locations—is growing. Flexible solar panels are lightweight and offer ultimate portability; however, charging mobile devices directly from flexible solar panels presents a significant problem. Most portable electronics are very sensitive to fluctuations in voltage and current and will cease charging if a fluctuation is sensed, which is likely to happen with a variable power source such as a solar panel.

To solve this problem, our senior design team designed a Mobile Power Station (MPS) which is essentially the missing link between the flexible solar panel and mobile electronic devices. The MPS supplies a constant source of power in order to charge mobile electronics, and it charges from a solar panel or constant power source, offering maximum flexibility. Cutting edge circuitry and algorithms were used to maximize MPS efficiency.

A need for the MPS was expressed by our client, PowerFilm, Inc., and will have applications in the military and commercial markets.

1.2 Acknowledgements

Our group would like to recognize and thank Brad Scandrett and Frank Jeffrey of PowerFilm, Inc. for answering all of our questions in a timely manner and working with us to develop the best possible product. Their knowledge of the solar market and solar products has been essential in developing the MPS.

We would also like to thank our advisor Dr. Fayed for his support in our senior design project. His industry knowledge and past experience has been a pivotal part of our design process. His insight has allowed us to break down our problems and develop efficient solutions.

Texas Instruments also deserves great recognition through their help with the TI Analog Design Contest. TI graciously provided us with all the parts and evaluation modules necessary to make this project a success.

Lastly, we'd like to thank Dr. Qiao, the EcPE senior design team, the Industry Review Panel, and CSG. They provide us with tools, help us utilize our resources, and challenge us to think about our problems from various aspects.

Our group appreciates the help and support of everyone that works diligently to provide the best possible senior design experience.

1.3 Problem Statement

As technology becomes more mobile, the power sources required to operate these devices must also become mobile. Photovoltaic solar arrays can be used to power mobile devices; however, solar array output power is non-constant and creates problems for mobile devices that require constant power.

Most mobile chargers require constant voltage and current, and will completely shut down in the event of a fluctuation.

PowerFilm, Inc. manufactures flexible amorphous silicon solar arrays and has requested a design for a MPS that will allow users to charge and operate their mobile devices in remote locations using flexible solar arrays. The MPS must mitigate the problem of non-constant solar array output power.

1.4 Potential solution

We propose to build a MPS containing lithium ion batteries to store energy gathered by PowerFilm flexible solar arrays. The MPS will store 100 watts of power, providing constant voltage and current to connected devices (loads). The lithium ion batteries serve as a buffer and mitigate voltage and current fluctuations. The MPS will implement a maximum power point tracking (MPPT) algorithm to maximize solar input power, and will contain lithium ion battery charge balancing circuitry. In addition to solar array input power, the MPS will also have the capability to charge from an AC-DC power supply; however, it will not charge from solar and AC-DC simultaneously.

1.5 Operating Environment

The MPS will be used in extreme outdoor conditions, an operating temperature of -20°C - 60°C is expected. Given the small dimensions and weight requirements of the MPS, it will likely be carried in a backpack while charging from a flexible solar array situated on the outside of the backpack. The case should completely enclose the circuitry, except for the input and output power ports, to minimize debris entry.

1.6 Intended Users

PowerFilm has indicated that the primary MPS user will be the United States military; however, PowerFilm also indicated interest in commercial sale of the MPS. The MPS was designed to provide power for an individual soldier during missions in remote locations. Commercial users may include outdoor enthusiasts and people requiring power in isolated locations.

1.7 Intended Uses

The MPS is intended to charge or power mobile devices such as laptops, cellular phones, iPods, and other mobile electronics. The MPS is designed to store 100 watts of power, which is intended to be enough energy to fully charge an “average” laptop battery. The main function of the MPS is to take energy collected from flexible solar arrays and use that energy to power mobile devices. The MPS is not intended to power large devices such as televisions or inductive appliances.

1.9 Market Survey

The MPS was designed upon the needs and specifications of PowerFilm Inc., which were determined from previous market research. The primary user of the MPS is the military, with the possibility of releasing a commercial version of the MPS. No significant additional market research was required.

1.10 Intellectual Property and Legal Considerations

PowerFilm, Inc. and Iowa State University EcPE will have co-ownership of MPS technology including any patents. Algorithms designed for the MPPT and charge controller circuitry will be covered in assumptions. Several algorithms for exist for MPPT, but the MPS will require an adapted version of current MPPT algorithms.

2. Approach

2.1 Concept Sketch

The complete MPS prototype will be encased in a rectangular container made out of a light weight material. Aluminum and plastic are both viable options. The exact dimensions of the case depend solely upon the dimensions of the PCB. The concept version of the MPS will be incased in a 8x8x2” plastic enclosure. With these dimensions, the MPS will fit comfortably into any backpack or large pocket. The concept sketch below is the MPS with all optional connectors.

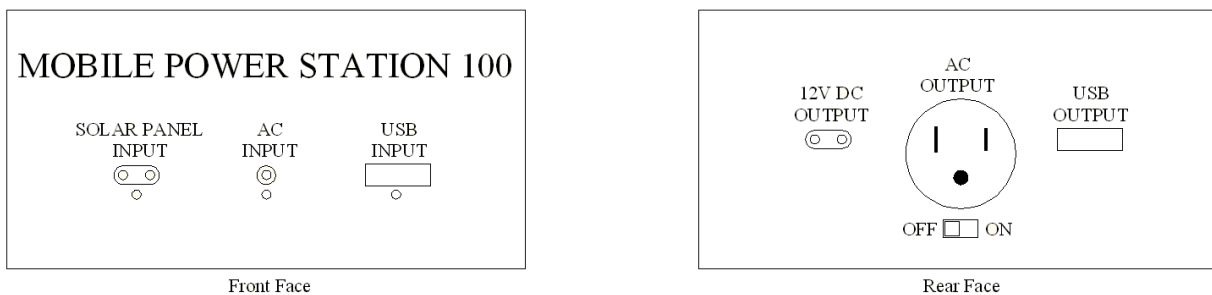


Figure 1: MPS Concept Sketch

2.2 System Block Diagram

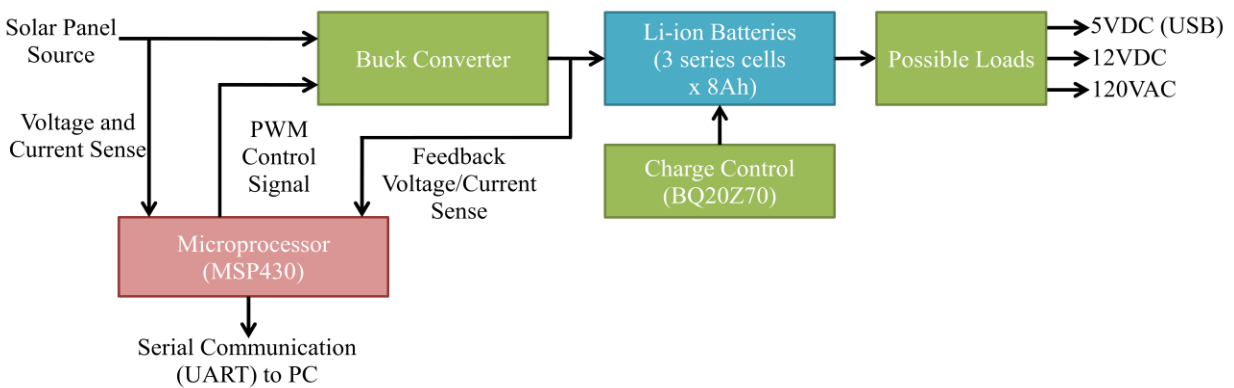


Figure 2: MPS Block Diagram

2.3 Functional Requirements

- FR 01 Optimized for standard solar panel input – 4A @ 15V (60W) Amorphous Silicon Panel
- FR 02 100W minimum Lithium-Ion battery capacity
- FR 03 15V DC input (with AC-DC Adapter)
- FR 04 12V DC output
- FR 05 Circuitry must be able to function in a temperature range of -20° C and 60°C
- FR 06 State of charge indicators
- FR 07 MPPT Charge controller with rating of up to 60W (4A @ 15V – Using 60W foldable on fully discharged batteries)
- FR 08 Charge Balancing Circuitry to keep Li-Ion Batteries balanced to prevent over or under charging
- FR 09 Temperature sensor for batteries with alarm LED
- FR 10 Achieve 80% or greater efficiency on all outputs

2.4 Non-functional Requirements

- NFR 01 The MPS shall be designed mainly for military soldier use
- NFR 02 The MPS should also be designed with options for commercial use
- NFR 03 The unit should have a weight of less than 5 pounds
- NFR 04 Unit should be manufactured for a cost of under \$500 per unit
- NFR 05 The unit should easily fit inside a military backpack

2.5 Assumptions and Limitations

Most technical assumptions and limitations are covered in the functional and non-functional requirements.

- Adequate senior design lab space will be provided.
- Adequate testing space and apparatus provided by PowerFilm, Inc.
- Evaluation boards for major components of the MPS (microprocessor & charging chip) are available
- PowerFilm, Inc. will provide solar panel and components not covered by the TI student program and senior design funding.
- PowerFilm, Inc. will design final the enclosure for the MPS circuitry. Our group needs only to build a prototype enclosure.
- Market research and requirements provided by PowerFilm, Inc. accurately represent the requirements of the MPS.
- The MPS will only be used as specified in intended uses and environments, as described in following sections.
- It will be more efficient to charge the MPS from a constant power supply if available.
- Technology developed by the senior design team will have co-ownership between PowerFilm, Inc. and Iowa State University EcPE.

2.6 Deliverables

Upon completion of the project, the team will provide:

- Project Plan
- Design Document
- PCB – Populated and tested for functional requirements
- Schematic diagram and PCB files
- Operational manual
- Final report

2.7 Safety Considerations

- Short Circuit
 - Fuses will be installed on the DC outputs to limit the output current to protect against electrical shorts
- Li-ion batteries
 - Batteries will be temperature monitored to keep batteries above -30°C and below 100°C to assure battery safety
 - Battery voltage should never exceed the limits of 2.0V lower limit and 4.2V upper limit
 - Current limits will be placed on the battery to assure that its charging or discharging rate never exceeds 2C
Note: Many Li-ion batteries can be charged and discharged at rates on the order of 10-20C, but for extended battery life and general cell stability, we will limit our outputs to 2C.
 - Charge time limited as secondary over charging protection

3 Design, Implementation & Testing

3.1 Design

3.1.1 Power Inputs

There are two power source inputs to the MPS: solar panel source, 15V AC/DC wall-wart adapter. The solar panel source is controlled by the MPPT algorithm to obtain maximum power transfer from the solar panel. The 120VAC source is converted to 15VDC by an external AC/DC converter.

The solar source is of great focus in this project, as it is based around MPPT. PowerFilm produces a variety of solar products that may be used with the MPS. PowerFilm has chosen to pair our MPS with its 20 Watt, 30 Watt and 60 Watt foldable solar chargers. PowerFilm also manufactures rollable solar chargers of varying wattages which also may be used. The current and voltage characteristics of these panels vary only slightly between designs and will not affect our design of the charger. Being paired with the 20, 30, and 60 watt foldable solar panels, our charger will in most cases see a solar panel charging current ranging from as low as 150mA (10% of 20W) and as much as 5 Amps. These currents were taken into account during the buck converter design and the Li-ion charging circuitry discussed later in this report.

3.1.2 MPPT

The characteristics of a solar cell are determined by the irradiation from the sun and the temperature of the panel, both of which change frequently. The MPPT adjusts the voltage across the solar panel to achieve maximum power from the panel. An algorithm preformed by a microprocessor will be used to find the voltage that generates maximum power.

The figure below shows a current vs. voltage curve of a typical solar panel. The smaller box inside the curve represents the maximum power output of the solar panel. Maximizing the area under the curve represents maximum power transfer from the solar array to the MPS.

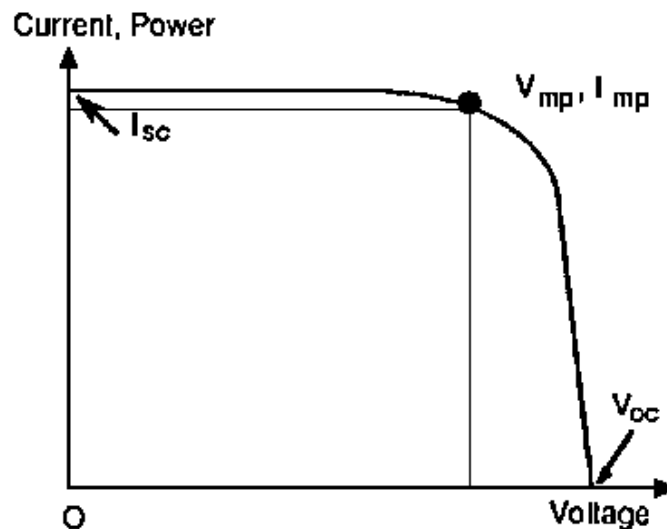


Figure 3: Current vs. Voltage Curve with Maximum Power Point [1]

MPPT is one of the key features of the MPS. We have chosen the Perturbation and Observation Method (POM) to find the Maximum Power Point (MPP) of the solar panel. The POM method works by imposing a voltage on the panel, calculating the power at the imposed voltage, then incrementing or decrementing the imposed voltage (by changing the PWM of the buck-converter) based on previous power measurements. This process can also be viewed as impedance matching between the solar panel's internal impedance and the load's impedance. Impedance matching is a common technique used in electrical engineering to achieve maximum power being delivered to the load. The details of the POM are shown in the figure below where $V(k)$, $I(k)$, and $p(k)$ ($p(k)=V(k) \times I(k)$) are the voltage, current, and power measured on the solar panel and V_{REF} is the voltage imposed on the panel. ΔV will be determined in the testing stage of our design, but will be in the range of 10's of mV.

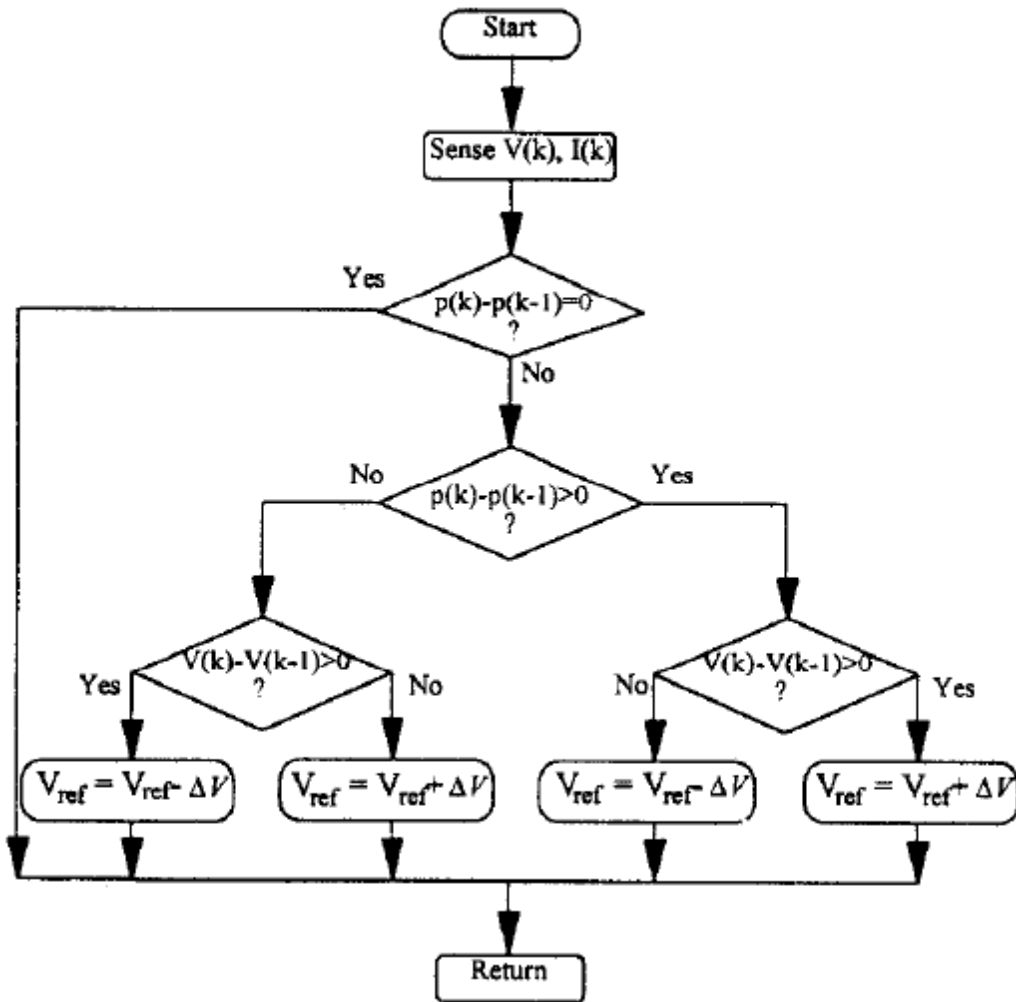


Figure 4: MPPT Perturbation and Observation Method [7]

3.1.2.1 Li-ion Charging Cycle

Li-ion battery charging cycle includes three phases: trickle, constant current, and constant voltage. The charging cycles must be strictly followed to prevent battery pack overcharge and fire. The MPS uses input and output voltage and current measurements to determine which charging phase the battery pack is in. The same measurements are also used for MPPT. MPPT is really only utilized during the constant current phase of charging, when the battery pack is demanding a large amount of current.

Battery pack voltage is most important measurement, and takes precedence over MPPT. The MPS utilizes a DC-DC buck converter making it very easy to hold an output voltage of 12.6V across the battery pack during constant voltage phase. The plot below best demonstrates the charging cycle.

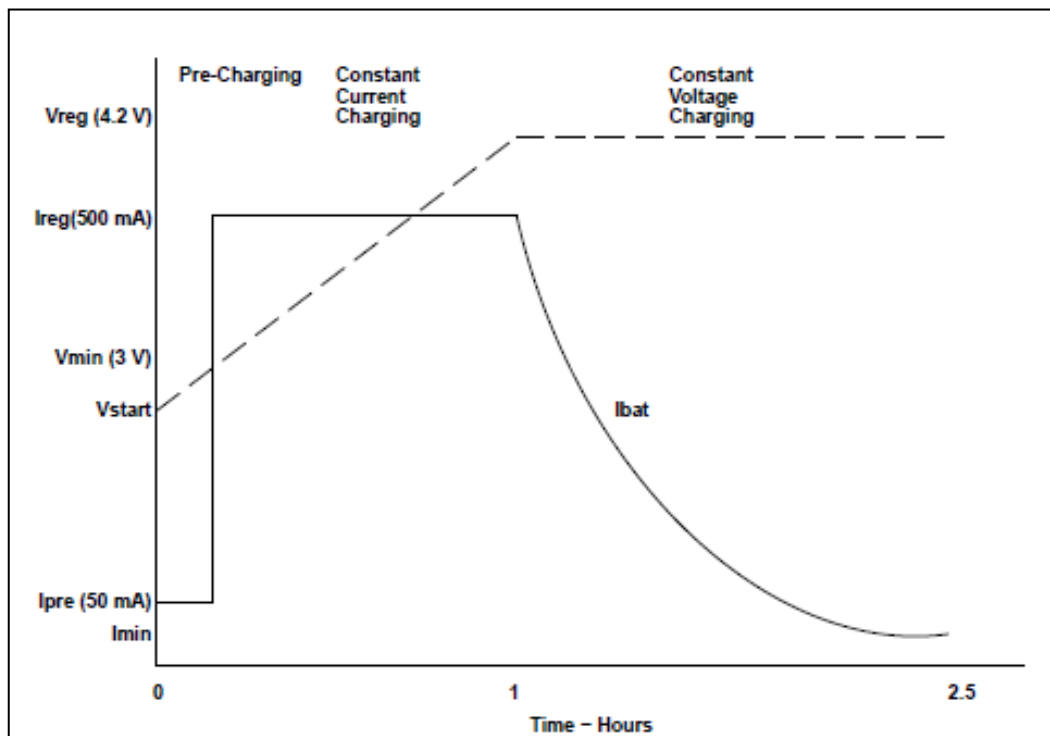


Figure 5: Li-ion Charging Cycle

3.1.3 Microprocessor

To accurately and efficiently find the MPP, two voltage measurements and one current measurement (which will actually be a voltage measurement across a current sense resistor) will need to be collected and stored within the microprocessor. The microprocessor must also generate a pulse-width modulated (PWM) signal, which will control the buck converter and, in turn, set the voltage across the solar panel. For testing purposes, the microprocessor must be able to communicate data back to a computer.

We chose TI's MSP430 Microprocessor for our microprocessor. Specifically, we will use the MSP430F2013 because it meets the requirements of our system: 3 analog-to-digital inputs, 1 timer to produce a PWM signal, and one additional output to be used for USART communication during the development stage of the MPS. Also, the MSP430F2013 comes in a compact 14-pin package and lacks unnecessary functionality. The figure below shows how each pin on the microprocessor will be implemented into our design.

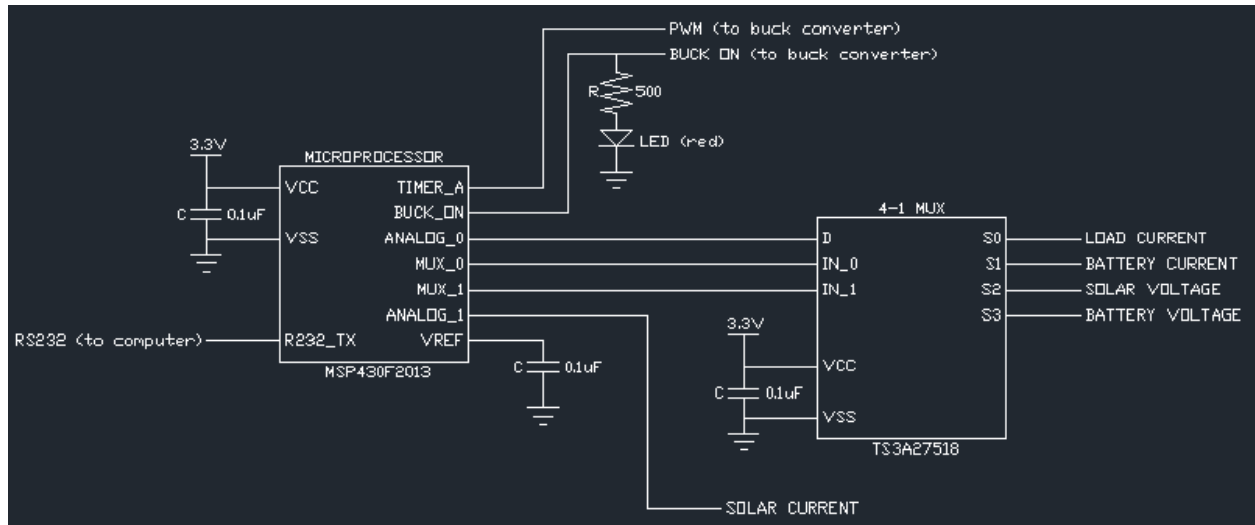


Figure 6: Microprocessor Schematic

3.1.3.1 VCC and VSS

The microprocessor will be indirectly powered from the solar panel. The microprocessor will share the same ground as the solar panel (VSS connects to the solar panel's negative terminal). A 3.3V linear regulator, connected directly to the solar panel, will step down the solar panel's voltage to 3.3V, which will power the microprocessor, multiplexor, and op-amps. When the solar panel is not generating power, the microprocessor will rest in an off state consuming no power from the MPS's Li-ion batteries. The instance the solar panel begins to generate power; the microprocessor will turn itself on and begin to run the MPPT algorithm.

3.1.3.2 Voltage Senses

The voltages across the solar panel and the battery/load are monitored while the MPS operates. The solar panel voltage, along with the solar panel current, is used to calculate the power generated by the solar panel. The battery/load voltage is used to determine which charging state the Li-ion batteries are in and stabilization during the constant voltage charging stage.

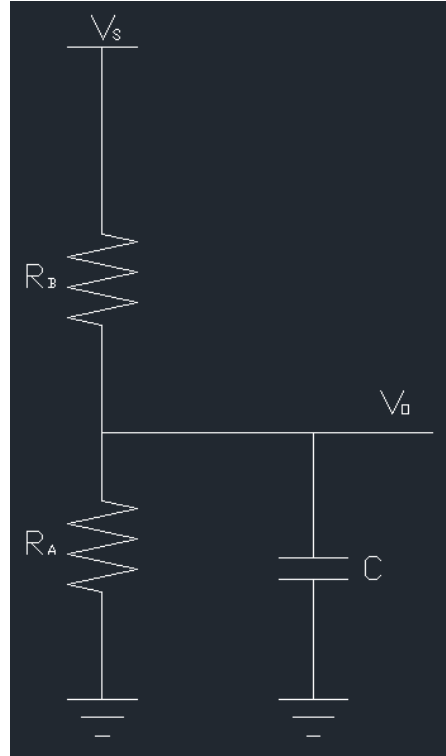


Figure 7: Voltage Sense Circuit

The voltage sense circuit consists of a simple voltage divider to step the voltage down to a level the microprocessor can read.

$$V_O = \frac{V_S * R_A}{R_A + R_B} = \frac{10k * V_S}{10k + 330k} = 0.029 * V_S V$$

$$Max\ reading = V_O = 0.6V, \quad V_S = 0.6/0.029 = 20.7 V$$

3.1.3.3 Low-side Current Sense

The current flowing through the battery and load are measured individually with low-side current sense circuits. The MPS uses the battery and load currents magnitude and direction to determine whether the battery is supplying power to the load or the buck converter is supplying power to the load. The MPS uses the battery current measurement for stabilization during the trickle charge stage of the Li-ion batteries.

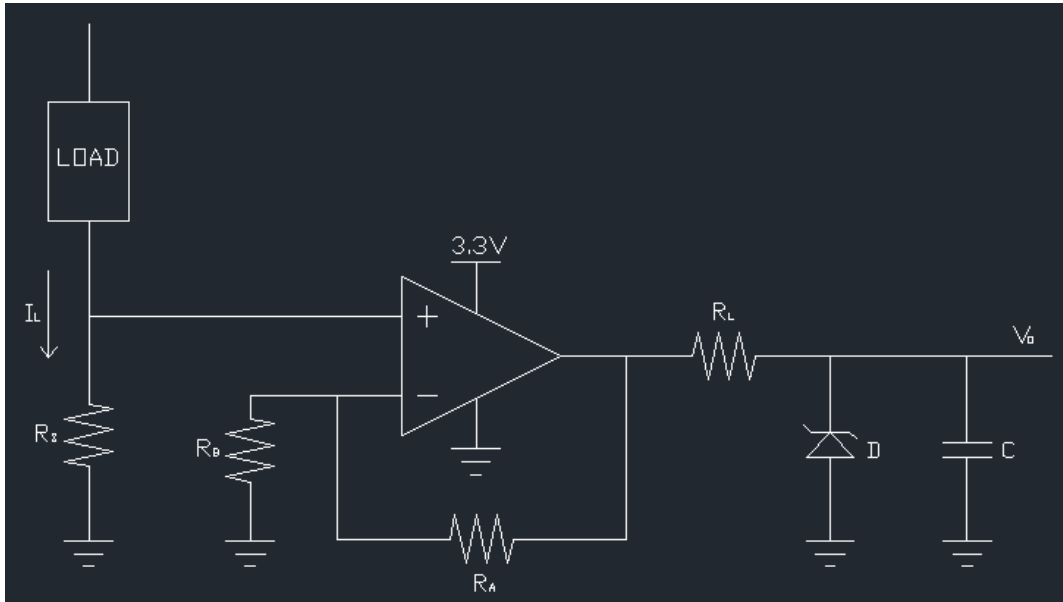


Figure 8: Low-side Current Sense Circuit

The low-side current sense circuit transforms the current (I_L) into a voltage which is amplified by the op-amp, set up in a non-inverting configuration, into a voltage the microprocessor can read. The relationship between the current (I_L) and output voltage (V_O) is shown in the equation below.

$$V_O = I_L * R_S * (1 + R_A/R_B) = .02 * (1 + 50k/1k) * I_L = 1.02 * I_L \text{ V}$$

$$\text{Max reading} = V_O = 0.6V, \quad I_L = 0.6/1.02 = 588 \text{ mA}$$

3.1.3.4 High-side Current Sense

In order to track the maximum power point, the current emitted from the solar panel needed to be measured. The only way to accomplish this is to use a high-side current sense circuit, which proved to be very difficult to implement. After trying two op-amps, we discovered an op-amp with rail-to-rail common-mode input capability was required because the common-mode voltage on the sense resistor is equal to op-amp's bias voltage. We chose Linear Technology's LT1677 precision op-amp for our design.

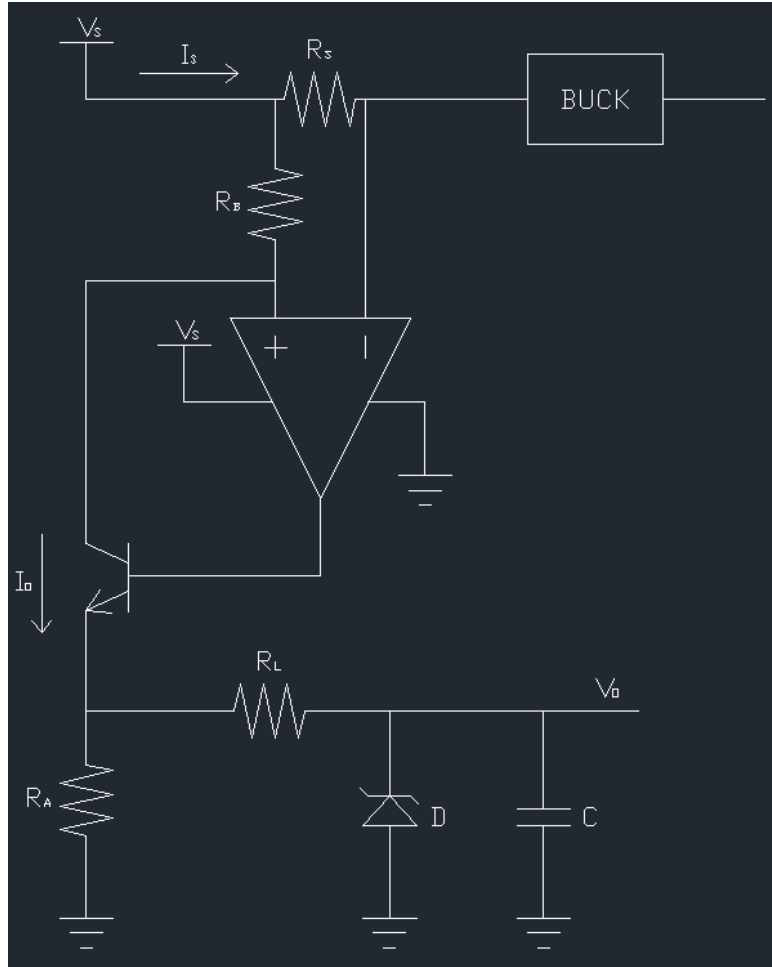


Figure 9: High-side Current Sense Circuit

The high-side current sense circuit turns the voltage across the sense resistor (R_s) into an output current (I_o). The output current is then converted into a voltage through R_A and read by the microprocessor. The relationship between the solar current (I_s) and output voltage (V_o) is shown in the equation below.

$$V_o = \frac{I_s * R_s * R_A}{R_B} = \frac{.02 * 120 * I_s}{20} = 0.12 * I_s V$$

$$Max\ reading = V_o = 0.6V, I_s = 0.6/1.2 = 5 A$$

3.1.3.5 UART Output

The MPS communicates serially using Universal Asynchronous Receiver/Transmitter (UART). The microprocessor did not come pre-programmed registers to easily communicate with a computer through a serial interface, therefore our team programmed the microprocessor to send bytes of data serially following the UART standard form of one start bit (logic low), 8 bits of data, and then two stop bits (logic high), at a Baud rate of 9600 bits/second.

To communicate with a computer through an RS-232 port, our system required a TTL/RS-232 driver. This driver transforms a logic low (0V) to +10V and a logic high (+3.3V) to -10V, which applies to RS-232 standards. This transformation is done through an IC chip (MAX232) which is shown below in a schematic is shown in Figure 7 (all capacitors are 1 μ F).

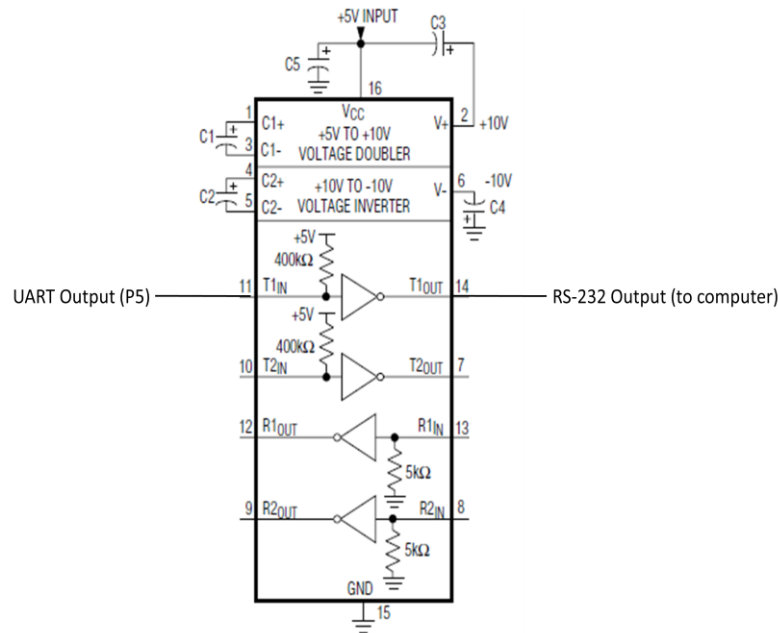


Figure 10: TTL/RS-232 Driver

3.1.4 Charging Circuitry

The charging circuitry controls which power input provides the MPS with energy. Only one power input will be selected at a time and a priority system is in place to select the source. The AC source is of highest priority (i.e. if the AC source is connected, it will power the MPS), and the solar source is of lowest priority (i.e. the solar source only powers the MPS if no other power input is connected). The charging circuitry will also turn on an LED next to the power input indicating which power input is charging the MPS. The charging circuitry ensures the batteries are charged evenly and are not under or overcharged, which could lead to catastrophic battery failure or battery damage.

The Charging chip used for our design is the BQ20Z70 Li-Ion charging IC made by Texas Instruments. Initially, the BQ78PL114 chip by Texas Instruments was evaluated. This was recommended by the faculty advisor, as it had been used on previous projects. It was found that the BQ78PL114 was too costly and complicated for our project and an alternative was chosen.

The secondary protection used in the MPS is a BQ20Z70. It is designed using TI's Impedance Track technology which effectively tracks the impedance of each battery to balance the batteries.

3.1.5 Batteries

The batteries store the energy delivered by the power inputs. They allow the MPS to supply power to the load without a power input available, which is limited by the power carrying capacity of the batteries. For this design we have chosen to use Lithium polymer batteries because of their high energy density and low profile design. Lithium polymer batteries are also capable of delivering much higher currents than lithium-ion which is important for possible high current loads especially through the AC output.

The batteries will be High Rate Lithium Polymer batteries with a nominal voltage of 3.7 volts per cell at 8 Amp hours per cell. We will use 3 batteries in series to provide a total internal nominal battery voltage of 11.1 Volts (12.6 Fully Charged).

3.1.6 Buck Converter

Our buck converter design utilizes N-FET's (N-Channel MOSFETs) in a two-phase configuration run by a MOSFET driver chip (IR 2104). This chip uses a charge pump to achieve the high voltages needed to switch the N-MOSFET. The driver chip is capable of accepting a PWM input from the Microprocessor and switches the MOSFETs accordingly. The Microprocessor can also control the operation of the buck converter using the "Shutdown" pin on the driver chip. This is useful in situations that may become unstable, in which the microprocessor can essentially shut down the entire system.

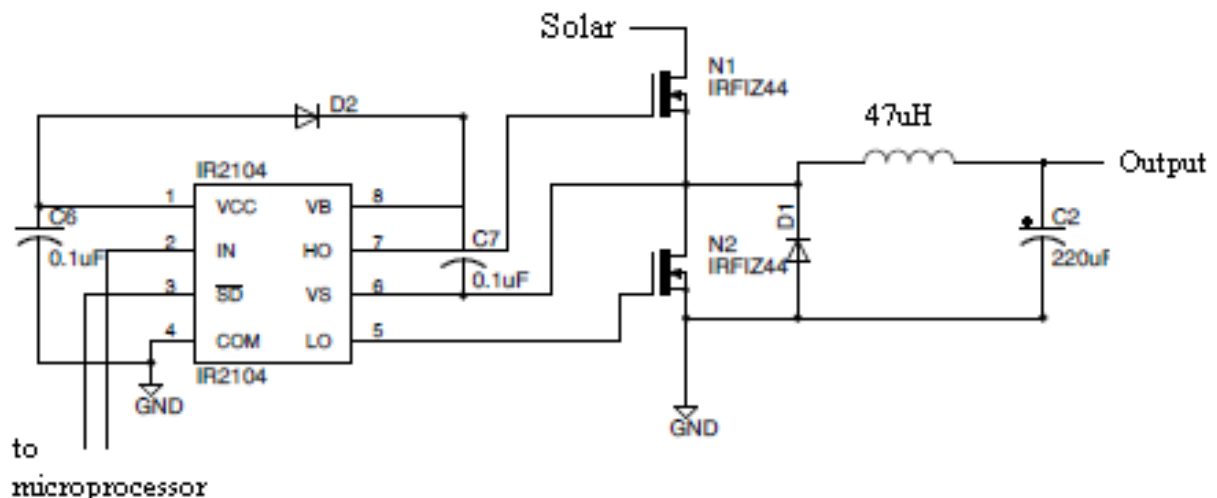


Figure 11: Buck Converter Schematic

To maximize efficiency we used the following equations to calculate ripple current and ripple voltage.

$$\text{Ripple Current: } \Delta I = \frac{(V_{in} - V_{out}) * V_{out}}{L * f * V_{in}}$$

$$\text{Ripple Voltage: } \Delta V = \frac{\Delta I}{8 * f * C}$$

In research on the type of solar panels that would be provided or packaged with our system, we needed to choose a ripple current that would leave our charger in Constant Current Mode (CCM) for almost all charging situations. Our design requires the converter to operate in CCM and not in Discontinuous Current Mode (DCM) because of the losses that our system would see during DCM mode. For this system, a design was chosen with a current ripple of less than 10% of our full charging current (load current). For a current ripple of less than 500mA, we found the closest popular inductor value to be 47uH. Therefore, the ripple current equation with the values is the following;

$$\Delta I = \frac{(V_{in} - V_{out}) * V_{out}}{L * f * V_{in}} = \frac{(15 - 12.6) * 12.6}{47 \times 10^{-6} * 100,000 * 15} = 0.429 \text{ Amps}$$

Using this ripple current, we can then calculate the output voltage ripple. We wanted to minimize our output ripple voltage to assure that there could be no inaccurate readings from the voltage feedback. For this, an output ripple voltage of 2-3mV was desired without the ESR considerations. A common 220uF capacitor size was chosen to provide the desired voltage ripple.

$$\Delta V = \frac{\Delta I}{8 * f * C} = \frac{0.429}{8 * 100000 * 220 \times 10^{-6}} = 0.0024 \text{ Volts}$$

With these values chosen and optimized for our design we assure that we can provide maximum charging efficiency.

Other efficiency considerations in the design include the input N-FET to the buck converter acting as a diode to prevent feedback current when no load is present.

3.2 Implementation

3.2.1 Initial Buck Converter

The initial design of the buck converter was implemented in the latter part of Senior Design I. Using sample parts and a breadboard, the buck converter was built. This tested the concept that our two phase NFET design would work as a buck converter. Using a function generator to supply the PWM signal and a power supply for both 3.3 and 15 volts, it was confirmed that the design would work as a buck converter. Duty cycle was adjusted to observe how output voltage was affected. With the buck converter working, soon we would be able to merge the microprocessor with the buck converter

3.2.2 Configuring Microprocessor to Control Buck Converter

With both main components of the MPS completed in implementation, the next step was to implement them into an entire system. This consisted of merging all the components onto one breadboard with a shared ground. Initial testing went fairly smoothly with a few bugs in the analog and digital conversions. We found that too much time was spent in the Interrupts of the microprocessor. This caused the PWM to periodically shut down, and thus shutting down our system. A MUX had to be implemented into the circuit to handle the number of measurements that had to be made.

Initially, there was a problem with our output voltage dropping out at high duty cycle but after consulting with our advisor we learned that because of the limitations of the MOSFET switching times, there was a ceiling on our maximum duty cycle. More testing was conducted and an upper limit of 93% duty cycle.

With everything implemented and running on a single breadboard, more extensive testing was conducted. Using an oscilloscope our group observed large transients (as high as 1 Volt p-p) at the output of our buck converter. These transients occurred at the switching point of the MOSFET controlled by the PWM signal.

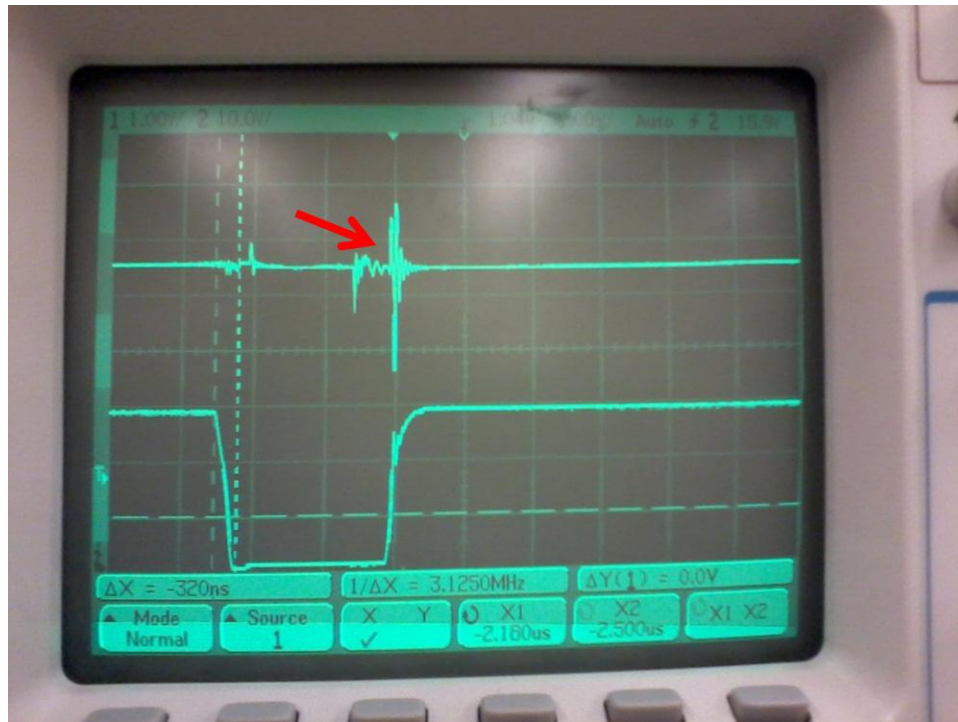


Figure 12: Large Transients observed at output of breadboard circuit

The transients had a proportional effect with current, i.e. as current increased, so did the magnitude of the voltage transient. Many weeks were spent on this problem because the importance of a fixed voltage for the lithium-ion batteries. Many modifications were made which included slowing the switch-ON of the MOSFET using snubber networks and applying extra filters to the high side MOSFET drain. A lower value inductor was used to attempt to essentially slow down the switching. During this stage, our group learned a great deal from research and our advisor about the many non-idealities of MOSFETs and our buck converter design. It was decided that the majority of the disturbance was due to the parasitic inductances and ESR (Equivalent Series Resistance) of the components and the breadboard. A combination of the described remedies was used to lower this transient to around 200mV.

During the month that was spent on this problem, our team also experienced difficulty with getting accurate voltage and current measurements. The voltage measurements were slightly inaccurate and floating, however our current measurements did not seem steady at all. We were not sure where this was coming from but our initial assumption was that it might be caused by the transients that we were seeing. After discussing it with our faculty advisor and doing some testing, we discovered that we were having a differential grounding issue. We attempted to connect all of our grounds in a star connection to cut down on this differential. The method helped slightly, however the problem still persisted. It was also found that the breadboard created a large resistance which was affecting our readings as well.

After a month of debugging on the breadboard we decided to finally implement our design onto a PCB. Our hope was that both the transients and the current measurements would be improved with a PCB design with a common ground plane and short traces.

3.2.3 Printed Circuit Board

In order to alleviate the voltage transients seen by the breadboard circuit, a PCB was designed. This design incorporated many of the design strategies that were learned through the process. Most important of which was the common ground plane with a star connection for the grounds.

The PCB schematic and board were made using Eagle CADsoft design software as per PowerFilm's suggestion and request. The design and layout was performed by Brad Jensen.

There were two revision of the PCB. The first PCB was a development PCB and didn't contain a solder mask, making modifications easy. We also discovered that we had selected the wrong differential amplifier for the high-side current sense. The amplifier we selected had an insufficient common-mode voltage input range. We selected and tested a new differential amplifier with a rail-to-rail common-mode voltage input, which effectively met out requirements.

There were a few small modifications that were implemented into PCB rev. 2 including: different high-side current sense differential amplifier, Zener diode protection, header pin consolidation, and a solder mask.

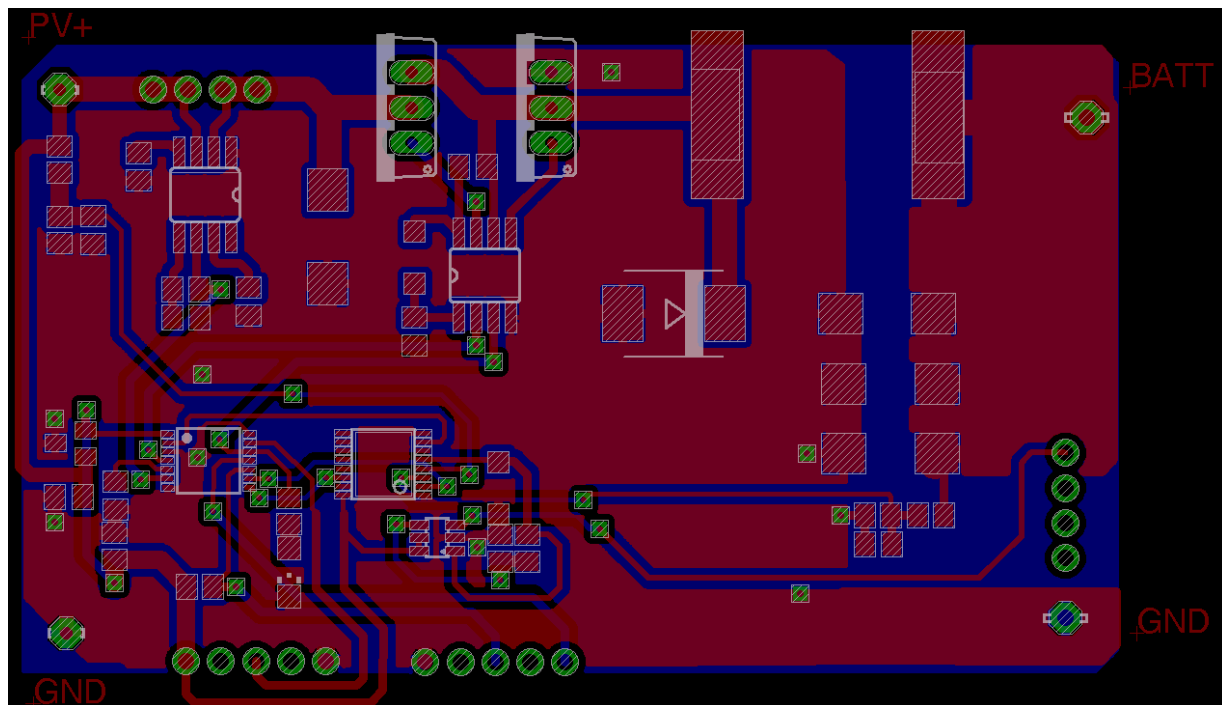


Figure 13: Prototype PCB

The PCB measured 3.7" x 2" and contained 39 components including resistors and capacitors. The prototype used two connectors for USART communication and for programming the microprocessor.

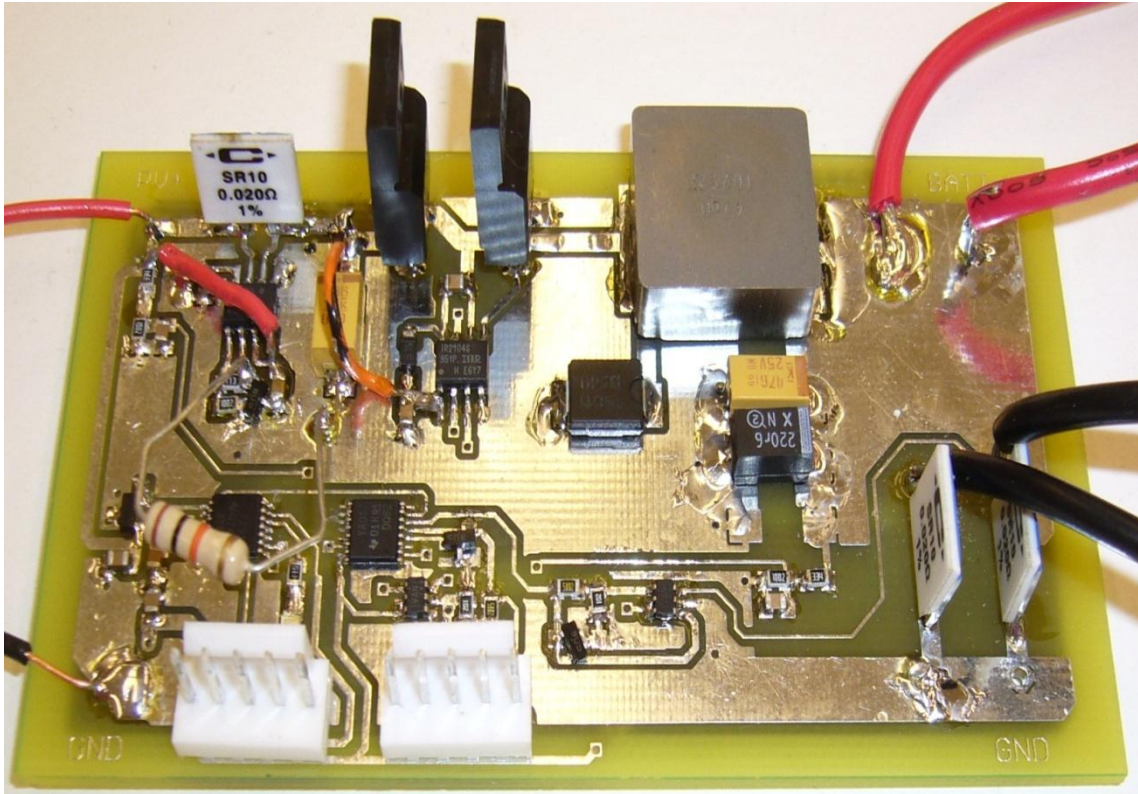


Figure 14: Populated PCB Rev. 1

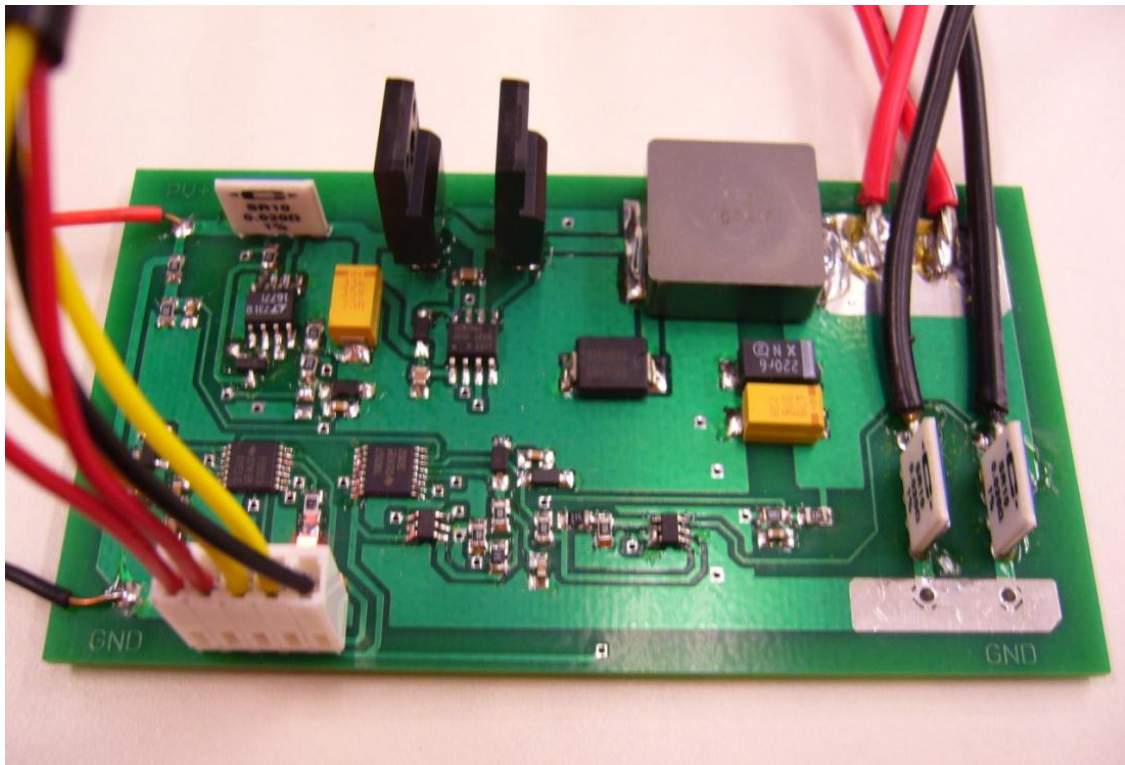


Figure 15: Populated PCB Rev. 2

3.3 Testing

3.1 PCB Tests

With the changes made onto a PCB, the entire system could be tested as a whole. The PCB was tested with a table power supply. Testing included measuring both Voltage and Current of the input and output using a resistive load. This information allowed for calibration of the voltage sensing, current sensing and overall efficiency of the system.

Next the resistive load was switched to a lead acid battery. The lead acid battery was used to test the charging algorithms. The lead acid battery is much safer to charger without any secondary protection. We observed a normal charging algorithm, however we did learn that at least in the case of the lead acid battery, the voltage sensing was slightly less accurate during charging. For this we decided that an averaging method was to be used inside the microprocessor to give more accurate readings.

3.2 Secondary Li-Ion Protection Testing

The safe charging of the li-ion batteries is definitely one of the most important parts of this project. The algorithm that is used in our project is supposed to safely charge the lithium ion batteries. As a practice in redundancy, a secondary protection/charge monitor is implemented on the output end of the buck converter. This charge monitor also needed to balance the series pack of batteries during the charging cycle.

Multiple charge cycles were run on the Li-Ion battery pack to test its performance. The evaluation board came with a software program that also helped in monitoring the battery characteristics as it charged.

The BQ20Z70 worked great for this application and will be recommended to PowerFilm for future applications

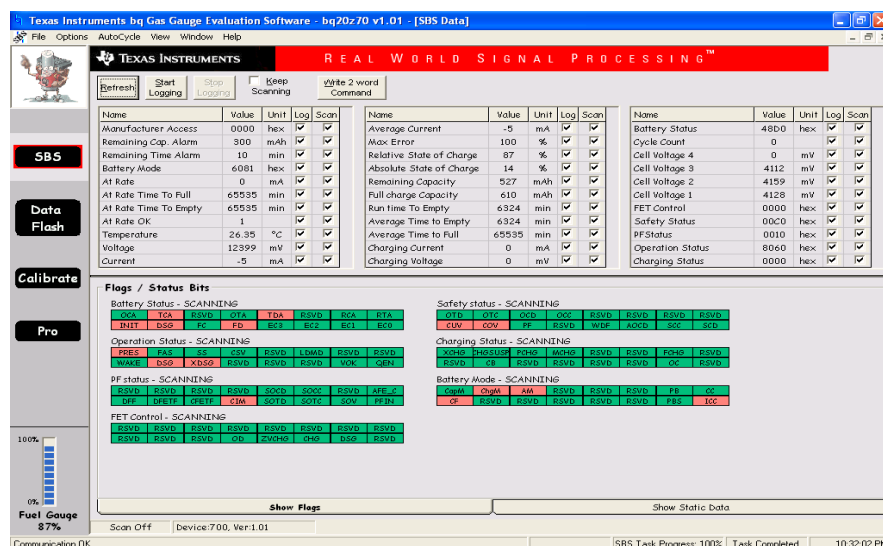


Figure 16: Software to Track and Monitor Cell Balancing

3.3 GUI Interface

A GUI interface was developed using “Processing,” a freeware Java GUI development program. The MPS GUI allowed us to quickly view and interpret the serial data being transmitted by the MPS430. Data such as input voltage, input current, battery voltage, battery current, load current, and duty cycle allowed us to check the charging algorithm. In addition to viewing data, the GUI also saved the data in CSV files that was easily imported to excel. Utilizing macros in Excel, we setup a calibration program to take the CSV file and turn the serial data from the MPS430 into readable voltage and current values.

The GUI also served as a great demonstration tool. Charging Li-ion batteries takes several hours, but users were able to see a dynamic plot of data and observe the current charging state (trickle, constant current, constant voltage, or off).

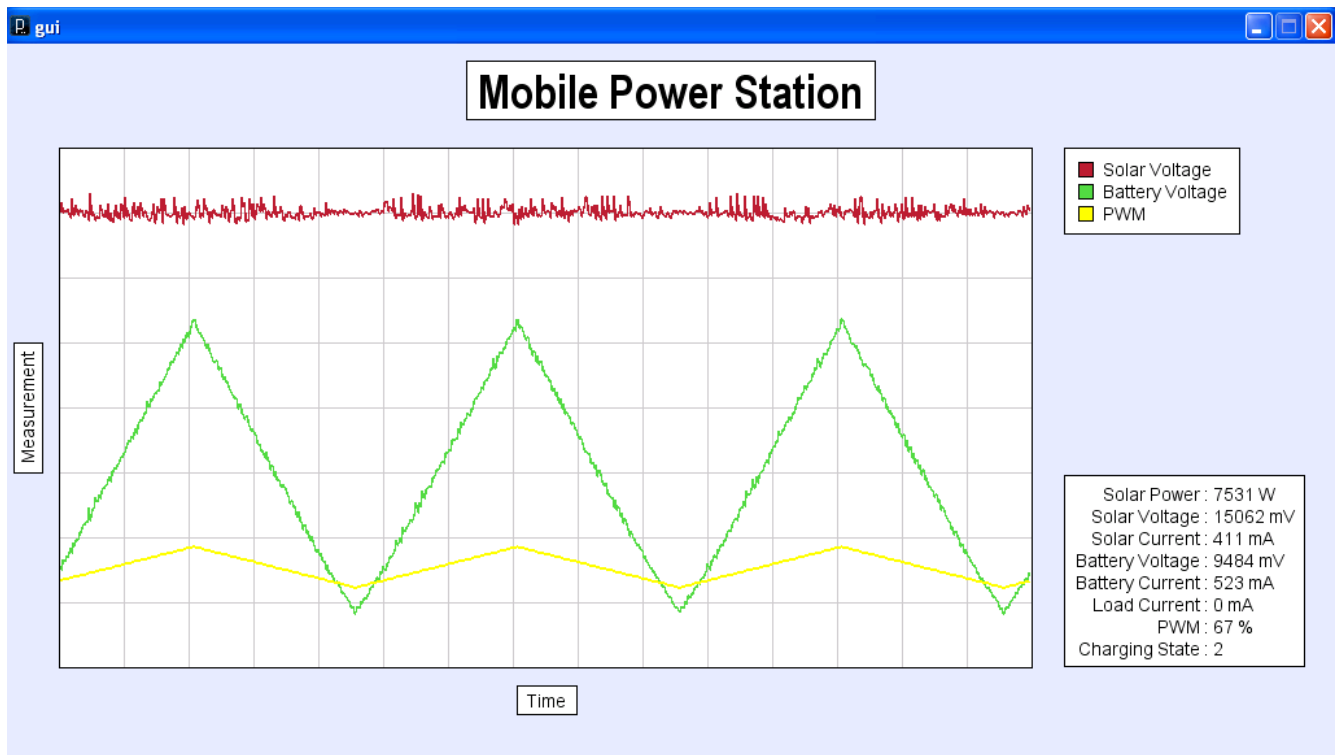


Figure 17: MPS GUI

3.4 Evaluation

One of the most important requirements of the MPS was to be efficient. As stated in functional requirement 08, the MPS should achieve 80% or greater efficiency. The entire goal of the MPS was to transfer as much solar energy to the battery pack as possible.

To measure efficiency, we tested several different load impedances and measured both input and output current to calculate input and output power. As you can see, we clearly achieved 80% efficiency. Testing was limited to strictly resistive loads. Further test should be performed with inductive loads.

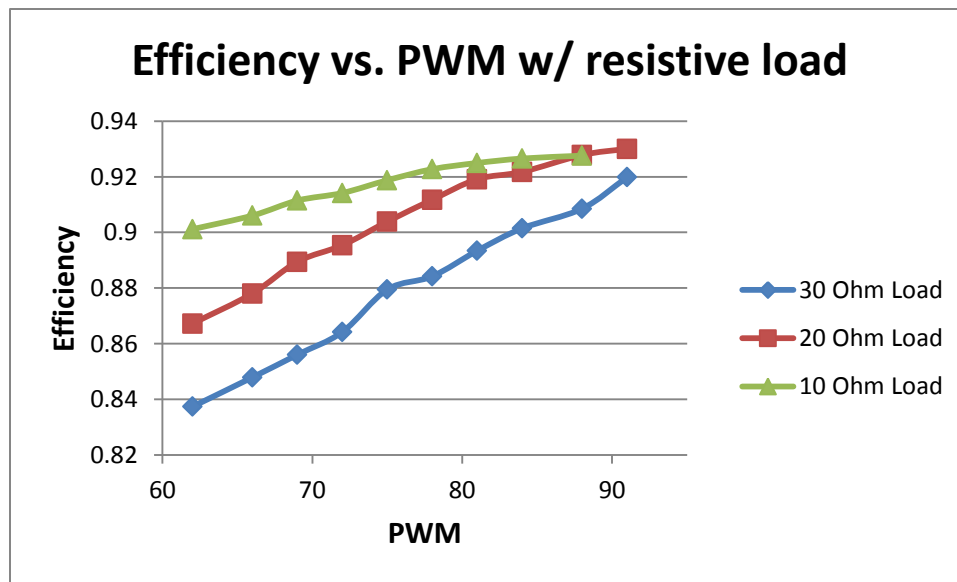


Figure 18: Efficiency vs. PWM

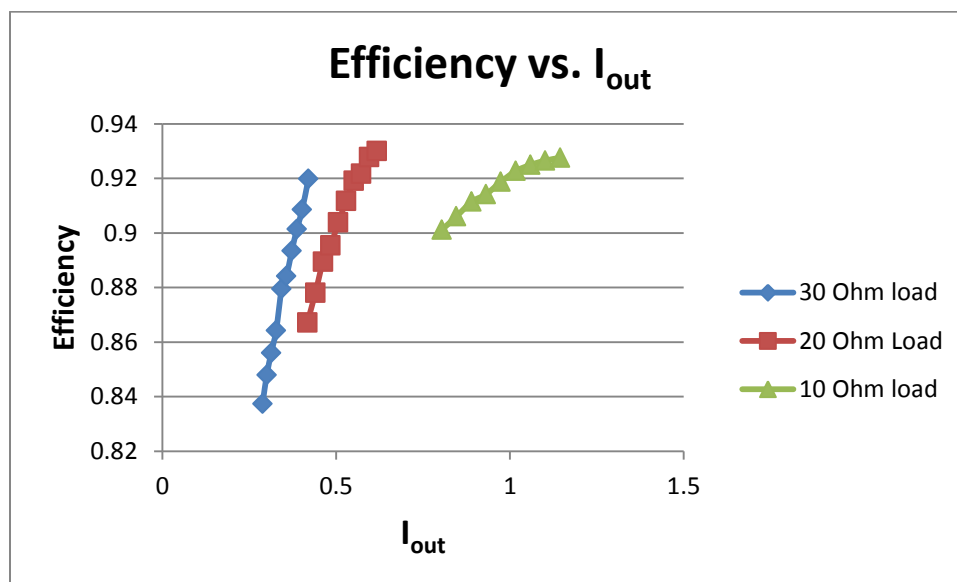


Figure 19: Efficiency vs. I_{out}

4 Resource Requirements

4.1 Team Utilization

An important part of any team is utilizing the talents of each member which is why we thought it was important to give a brief description of each team members' role and talents.

- Brad Jensen
 - Role: PCB Designer & Communication Liaison
 - Designs populates surface mount PCBs
 - Troubleshoots circuit problems
 - Coordinates with the PowerFilm, Inc. engineers
 - Relays relevant and important information between team, client, and advisor.
 - Talents and Contributions:
 - Works for PowerFilm, Inc.
 - PCB design and layout
 - Experience with battery management

Brad Jensen has been a great asset to the team. Brad spent the majority of his time on research and implementation of the buck converter circuitry. He produced the PCB layout which helped push the project forward. He has kept close communication with PowerFilm engineers on many of the design aspects. Brad has also collaborated with both Nathan Schares and William Klema to improve the design of the MPS.

- William Klema
 - Role: Circuit Designer & Webmaster
 - Circuit design, op-amp circuits
 - Designed and updated senior design webpage
 - Organize and archive project documents
 - Talents and Contributions:
 - VLSI circuit design
 - Code ninja

William Klema's experience with the programming aspect of the project has helped tremendously. He has also updated and maintained the website regularly with the weekly reports and other documents and presentations. William was able to program the MSP430 for USART communication and has worked on the ADC channel inputs on the MSP430. He now has a great handle on the MSP430's operation which is very important as the microcontroller is the heart of the MPS.

- Nathan Schares
 - Role: Project Manager
 - Plan meetings
 - Manage team tasks
 - Publish weekly reports
 - Publish group documents
 - Talents and Contributions:
 - Leadership
 - Project proposal experience
 - Power electronics and control systems

Nathan Schares leadership has helped guide our team in the right direction and kept us on track throughout the design process. Nathan dedicated much of his time to fully understanding the operation of the entire circuitry from the MPPT to the buck converter design. Nathan has also submitted our weekly reports and presentations on time. His dedication to the project has helped ensure its success.

We would also like to recognize Professor Ayman Fayed for his immense help through the design process. Professor Fayed was always willing and able to help answer any of our questions with great detail. He has been a wonderful resource to our team and much of our progress is due to his guidance and teaching.

4.2 Material list and estimates

Table 1 provides information regarding cost of electrical parts for the final product.

Electrical Parts Cost Per Unit Estimate	
Part/Material	Cost(\$)
Mobile Power Station Electrical Components (Senior Design)	21.25
Li-Ion External Battery Protection (TI circuitry)	9.55
PCB	5
Batteries	97.25
AC/DC Adapter Parts and Circuitry	12.75
Total	145.80

Table 1: Electrical Parts Cost Per Unit Estimate

4.3 Human Resource Estimates

Figure 10 provides estimates for labor requirements of each section of this project. These values are purely estimates.

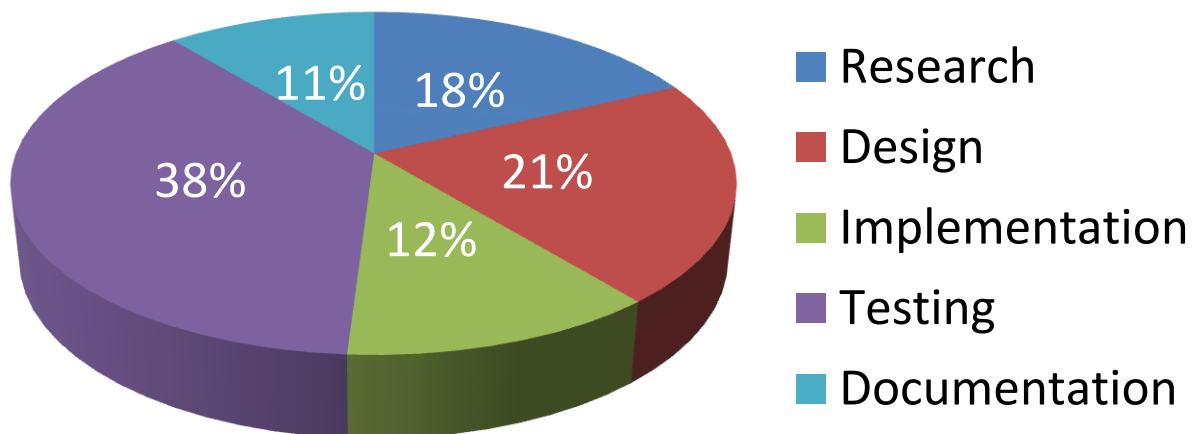


Figure 20 - Human Resources Pie Chart

4.4 Project Schedule

Below is the initial plan for the second semester of the project. Unfortunately we were not able to stick exactly to the schedule; however we were able to accomplish our goals within the desired time limits.

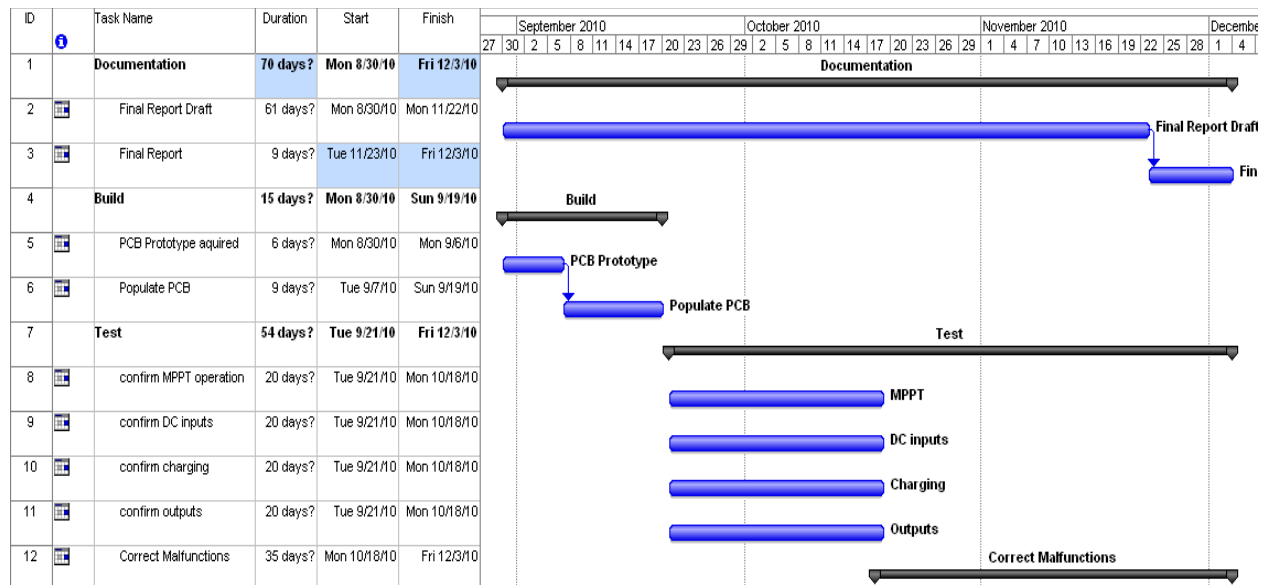


Figure 21: Semester 2 planned schedule

4.5 Deliverables

The team delivered a working prototype of the MPS to PowerFilm Solar at the end of the semester. All documentation files were included so that PowerFilm could continue work on the design. In addition to the prototype and data files, our group created a 32x40" poster, a 30 minute presentation, and this final report to summarize our work. A modified version of this report will also be sent to Texas Instruments, as part of the Analog Design Contest.

5 Project Team Information

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6. Closing Summary

The MPS is small, lightweight and robust, offering users a reliable source of solar energy in remote locations. The option of charging from a flexible solar panel or a constant power source adds versatility to the MPS's 100W battery capacity.

The team has learned an enormous amount about the design process through this class and project. From idea, to design, to implementation and testing, the team worked hard to meet the requirements of the project. Good use of critical thinking, time management, and hard work undoubtedly played a huge part in the success of this project.

References

- [1] "IV curve with MPPT." Australian National University. Web. 25 Feb. 2010.
<<http://engnet.anu.edu.au/DEpeople/Andres.Cuevas/Sun/help/PVguide.html>>.
- [2] Chapman, P.L., T.Esram. *Comparison of Photovoltaic Array Maximum Power Point Tracking Techniques*. 2005. Print.
- [3] Farando, Roberto, and Sonia Leva. *Energy comparison of MPPT techniques for PV Systems*. Tech. no. 1790-5060. Milan: Politecnico di Milano, 2008. Print.
- [4] Liu, C., B. Wu, and R. Cheung. *Advanced Algorithm for MPPT Control of Photovoltaic Systems*. Tech. Montreal: Canadian Solar Buildings Conference, 2004. Print.
- [5] Kaito, Koizumi, and Goshima. *Development of MPPT Algorithm for Digitally Controlled PV Inverter*. Tech. Tokyo University of Agriculture and Technology. Print.
- [6] "Large Capacity Li-Polymer Cells." Yoku Energy. Web. 25 Feb. 2010.
<<http://www.yokuenergy.com/en/proindex204.asp>>.
- [7] Hua, C., Shen, C., 1998. Study of maximum power tracking techniques and control of dc/dc converter for photovoltaic power system. In: Atsuo K. (Ed.), *Proceedings of the IEEE PESC, Fukuoka, Japan*, pp. 86–93.
- [8] Lujara, Nelson K., Jacobus D. Van Wyk, and Peter N. Materu. "IEEE." *Power Electronic Loss Models of Dc-dc Converters in Photovoltaic Applications* (1998). Web.
- [9] Nolan, Tim. "Arduino Solar Charger". 04/25/10
<<http://www.timnolan.com/index.php?page=arduino-ppt-solar-charger>>.

Appendix I – Microprocessor Code

```
#include "msp430x20x3.h"

#define PWM_MIN    64    //20% duty cycle
#define PWM_MAX    302   //95% duty cycle
#define A_100MAbat  1261  //100mA, tuned for PCB
#define A_100MAload 2506  //100mA, tuned for PCB
#define V_126V     9542   //12.6V, tuned for PCB
#define V_125V     9465   //12.5V, tuned for PCB
#define V_9V       6757   //9V, tuned for PCB

//Functions
void buck_control(unsigned int measurement, unsigned int desired);
void mppt(void);
void pwm(char direction);
unsigned int adc(char channel);
void wait_ms(unsigned int ms);
void display_data(void);
void transmit_int(unsigned int number);

//Global variables
unsigned int measurement = 0; //stores analog reading
char charging_state = 0;    // 0:trickle, 1:mppt, 2:constant voltage,
4:shutdown
char adc_complete = 0; //flag for adc measurement
char pwm_limit = 0;    //pwm limit reached counter, currently not in use
signed int state_advance = 0; //state advance counter
char pwm_up = 1; //direction of pwm, used for mppt
unsigned int bat_voltage = 0; //voltage of batter pack
unsigned int bat_current = 0; //current through battery pack
unsigned int load_current = 0; //current through load
unsigned int solar_voltage = 0; //voltage of solar cell
unsigned int solar_current = 0; //current through solar cell
unsigned long power_0 = 0; //previous power

//UART variables
unsigned int data = 0; // Variable for transmitted USART data
char usart_state = 0; // State variable for USART
char i = 0; // Bit counter for USART
char delay = 0; // Delay counter for USART
char totaldelay = 33; // Sets baud rate of 9680

int main(void)
{
    //Disables Watchdog Timer
    WDTCTL = WDTPW + WDTHOLD;

    //Set MCLK and SMCLK to 16MHz
    BCSCTL1 = CALBC1_16MHZ;
    DCOCTL = CALDCO_16MHZ;

    //SD16 setup (16-bit ADC)
    P1SEL |= 0x08; // Select P1.3 for voltage reference
    SD16CTL = SD16REFON + SD16SSEL_1 + 0x0400; // 1.2V ref, SMCLK/16
```

```

SD16INCTL0 = SD16INCH0;          // Set channel A0+/-
SD16CCTL0 |= SD16SNGL + SD16UNI + SD16IE;    // Single conv, 256OSR,
unipolar, enable interrupt
SD16AE = SD16AE0;                // Reset external input to A0+/-
SD16INCTL0 = SD16INCH_0;        // Reset channel observe

//USI setup (UART)
P2SEL = 0x00;                    // Sets Port 2 to I/O
P2DIR |= 0x40;                   // Set P2.6 to output direction
USICTL0 |= USIMST;               // Master mode
USICTL1 |= USIIE;                // Counter interrupt, flag
remains set
USICKCTL = USIDIV_0 + USISSEL_2; // divide by 1, SMCLK
USICTL0 &= ~USISWRST;           // USI released for operation

//MUX Control
P1DIR |= 0xC0; //Set P1.6, P1.7 to output direction for MUX control

//Buck control/LED
P2DIR |= 0x80; //Set P2.7 to output direction
P2OUT &= 0x7F; //LED off

//Timer1 PWM setup
P1DIR |= 0x04;                    // Set P1.2 to output
P1SEL |= 0x04;                    // Select P1.2 for TA1 output
TACCTL1 = OUTMOD_7;              // TACCR1 reset/set, interput enable
TACTL = TASSEL_2 + MC_1;         // SMCLK, upmode
TACCR0 = 318;                    // PWM frequency 50kHz
TACCR1 = 0;                      // Duty cycle, min 95(60%), max 143(90%)

_BIS_SR(GIE); //Enable interrupts

//determines starting state
bat_voltage = adc(4); //reads battery voltage
if(bat_voltage < V_9V) charging_state = 0; //trickle charge
else if(bat_voltage < V_126V) charging_state = 1; //mppt
else charging_state = 2; //constant voltage

wait_ms(14000); //testing purposes

P2OUT |= 0x80; //turns on buck/LED
TACCR1 = PWM_MAX;
wait_ms(100); //100ms delay to allow buck converter to settle

/*
if(sweep==1) {
    do{
        for(TACCR1=250; TACCR1<=300; TACCR1+=5) {
            for(int n=0; n<10; n++) {
                for(int i=0;i<=100;i++) {
                    for(int j=0;j<=510;j++) {
                        }
                    }
                }
            }
        }
        //wait_ms(100);
        load_current = adc(0);
        bat_current = adc(1);
        solar_voltage = adc(2);
    }
}

```

```

        bat_voltage = adc(4);
        solar_current = adc(8);
        display_data();
    }
}
for(TACCR1=140; TACCR1>100; TACCR1-=5) {
    for(int n=0; n<10; n++) {
        wait_ms(500);
        load_current = adc(0);
        bat_current = adc(1);
        solar_voltage = adc(2);
        bat_voltage = adc(4);
        solar_current = adc(8);
        display_data();
    }
}
} while(1);
}*/

//Program loop
while(1) {

    switch(charging_state) {

        case 0: //trickle charge, keep battery current at 100mA
            bat_current = adc(1); //reads battery current
            buck_control(bat_current, A_100MAbat); //seeks 100mA
            bat_voltage = adc(4); //reads battery voltage
            if(bat_voltage >= V_9V) state_advance++; //bat_voltage is greater
than 9V
            else state_advance=0;
            break;

        case 1: //mppt
            mppt(); //finds the mpp
            bat_voltage = adc(4); //reads battery voltage
            if(bat_voltage >= V_126V) state_advance++; //bat_voltage is greater
than 9V
            else state_advance=0;
            break;

        case 2: //constant voltage
            bat_voltage = adc(4); //reads battery voltage
            buck_control(bat_voltage, V_126V); //seeks 12.6V
            bat_current = adc(1); //reads battery current
            load_current = adc(0); //reads load current
            if(bat_current <= A_100MAbat && load_current <= A_100MAload)
state_advance++; //bat_current is less than 50ma
            else if(bat_voltage < V_125V) state_advance--;
            else state_advance=0;
            break;

        case 4: //shutdown
            pwm_limit = 0; //resets pwm limit counter
            P2OUT &= 0x7F; //turns buck/LED off
            bat_voltage = adc(4); //reads battery voltage

```

```

        if(bat_voltage < V_126V) state_advance++; //bat_voltage is less than
12.6V
        else state_advance=0;
        break;
    }

    if(state_advance>10) { //advance to next state
        charging_state++;
        if(charging_state==3) charging_state++;
        else if(charging_state==5) {
            charging_state=0;
            P2OUT |= 0x80; //turns on buck/LED
        }
        state_advance=0;
    }
    else if(state_advance<-10) {
        charging_state=1;
        state_advance=0;
    }

    //testing and presentation purposes
    load_current = adc(0);
    bat_current = adc(1);
    solar_voltage = adc(2);
    bat_voltage = adc(4);
    solar_current = adc(8);
    display_data();

} //end program while loop
} //end main

//Brings buck converter towards specified voltage
void buck_control(unsigned int measurement, unsigned int desired) {

    signed int delta = measurement - desired;

    if(delta > 0) pwm(0); //buck voltage too high
    else if(delta < 0) pwm(1); //buck voltage too low
    else pwm_limit = 0; //buck voltage right on

    wait_ms(10); //give buck converter time to settle
}

//Seeks maximum power point
void mppt(void) {
    long power_1;

    solar_voltage = adc(2);
    solar_current = adc(8);
    power_1 = (long)solar_voltage * solar_current; //current power

    if(pwm_up == 1) { //increasing pwm
        if(power_1 > power_0) pwm(1); //increase pwm
        else {
            pwm(0); //decrease pwm

```

```

        pwm_up = 0;
    }
}
else { //decreasing buck voltage
    if(power_1 > power_0) pwm(0); //decrease pwm
    else {
        pwm(1); //increase pwm
        pwm_up = 1;
    }
}

power_0 = power_1;
}
//Adjusts PWM duty cycle
void pwm(char direction) {

    if(direction==1) {
        if(TACCR1 < PWM_MAX) { //checks PWM max value
            TACCR1 += 1;
            pwm_limit = 0;
        }
        else pwm_limit++;
    }
    else {
        if(TACCR1 > PWM_MIN) { //checks PWM min value
            TACCR1 -= 1;
            pwm_limit = 0;
        }
        else pwm_limit++;
    }

    //if(pwm_limit >= 100) charging_state = 4; //desired current cannot be
    reached...shutdown
}

//Reads specified analog voltage
//Takes average of 40 readings and returns results
unsigned int adc(char channel) {

    unsigned int measurements = 0;
    unsigned long average = 0;
    int trials = 10;

    switch(channel) {
        case 0:
            P1OUT &= 0x3F; //load_current, in2=0, in1=0
            break;
        case 1:
            P1OUT &=0x7F; P1OUT |= 0x40; //bat_current, in2=0, in1=1
            break;
        case 2:
            P1OUT |=0x80; P1OUT &= 0xBF; //solar_voltage, in2=1, in1=0
            break;
        case 4:
            P1OUT |= 0xC0; //bat_voltage, in2=1, in1=1
            break;
    }
}

```

```

    case 8: //solar_current, A2
        SD16AE &= ~SD16AE0;           // Disable external input A0+, A0
        SD16INCTL0 &= ~SD16INCH_0;    // Disable channel A0+/-

        SD16INCTL0 |= SD16INCH_2;     // Enable channel A2+/-
        SD16AE |= SD16AE4;            // Enable external input on A2+
        trials = 50;
        break;
    }

    for(int m=0; m<trials; m++) {
        for(int a=0; a<4; a++) {
            SD16CCTL0 |= SD16SC; //starts conversion
            while(adc_complete==0); //waits til conversion is complete
            measurements += measurement/4;
            adc_complete=0;
        }
        measurements /= 4;
        average += measurements;
        measurements = 0;
    }
    average /= trials;

    if(channel==8) {
        SD16AE = SD16AE0;           // Reset external input to A0+/-
        SD16INCTL0 = SD16INCH_0;    // Reset channel observe
        average+=(11269-solar_voltage)*60/100;
        if(average>16384) average=1;
    }

    return (int)average;
}

//SR16 (16-bit ADC) interrupt service routine
//Loop initiates once whenever SD16SC = 1
#pragma vector = SD16_VECTOR
__interrupt void SD16ISR(void)
{
    measurement = SD16MEM0;
    adc_complete = 1;
}

//Delays program
void wait_ms(unsigned int ms)
{
    unsigned int i,j;
    for(i=0;i<=ms;i++) {
        for(j=0;j<=510;j++) {
        }
    }
}

//Sends all data through serial port
void display_data(void)
{
    unsigned int n=0;

```



```

while(n<65000) {
    while(usart_state!=0); //waits until usart is ready
    switch(n) {
        case 0:
            data = 0x000B; //new line
            usart_state = 1; //begin transmit
            n=1;
            break;
        case 1:
            data = 0x000D; //carriage return
            usart_state = 1; //begin transmit
            n=2;
            break;
        case 2:
            transmit_int(load_current);
            usart_state = 1; //begin transmit
            n=4;
            break;
        case 4:
            data = 0x0009; //tab
            usart_state = 1; //begin transmit
            n=8;
            break;
        case 8:
            transmit_int(bat_current);
            n=16;
            break;
        case 16:
            data = 0x0009; //tab
            usart_state = 1; //begin transmit
            n=32;
            break;
        case 32:
            transmit_int(solar_voltage);
            n=64;
            break;
        case 64:
            data = 0x0009; //tab
            usart_state = 1; //begin transmit
            n=128;
            break;
        case 128:
            transmit_int(bat_voltage);
            n=256;
            break;
        case 256:
            data = 0x0009; //tab
            usart_state = 1; //begin transmit
            n=512;
            break;
        case 512:
            transmit_int(solar_current);
            n=1028;
            break;
        case 1028:
            data = 0x0009; //tab
            usart_state = 1; //begin transmit

```

```

        n=2056;
    break;
    case 2056:
        transmit_int(state_advance);
        n=4112;
    break;
    case 4112:
        data = 0x0009; //tab
        usart_state = 1; //begin transmit
        n=8224;
    break;
    case 8224:
        transmit_int(charging_state);
        n=16448;
    break;
    case 16448:
        data = 0x0009; //tab
        usart_state = 1; //begin transmit
        n=32896;
    break;
    case 32896:
        transmit_int(TACCR1);
        n=65000;
    break;
}
}
}

//Breaks down an integer into individual characters and transmits them via
UART
void transmit_int(unsigned int number)
{
    while(usart_state!=0);
    data = ((number/10000)%10) | 0x0030;
    usart_state = 1;

    while(usart_state!=0);
    data = ((number/1000)%10) | 0x0030;
    usart_state = 1;

    while(usart_state!=0);
    data = ((number/100)%10) | 0x0030;
    usart_state = 1;

    while(usart_state!=0);
    data = ((number/10)%10) | 0x0030;
    usart_state = 1;

    while(usart_state!=0);
    data = (number%10) | 0x0030;
    usart_state = 1;
}

//USI interrupt service routine
//Transmits data serially through P2.6 following UART standards
//Baudrate 9600 bits/sec, 2 stop bits, no parity

```

```

#pragma vector=USI_VECTOR
__interrupt void universal_serial_interface(void)
{
    USICNT = 0x1F;

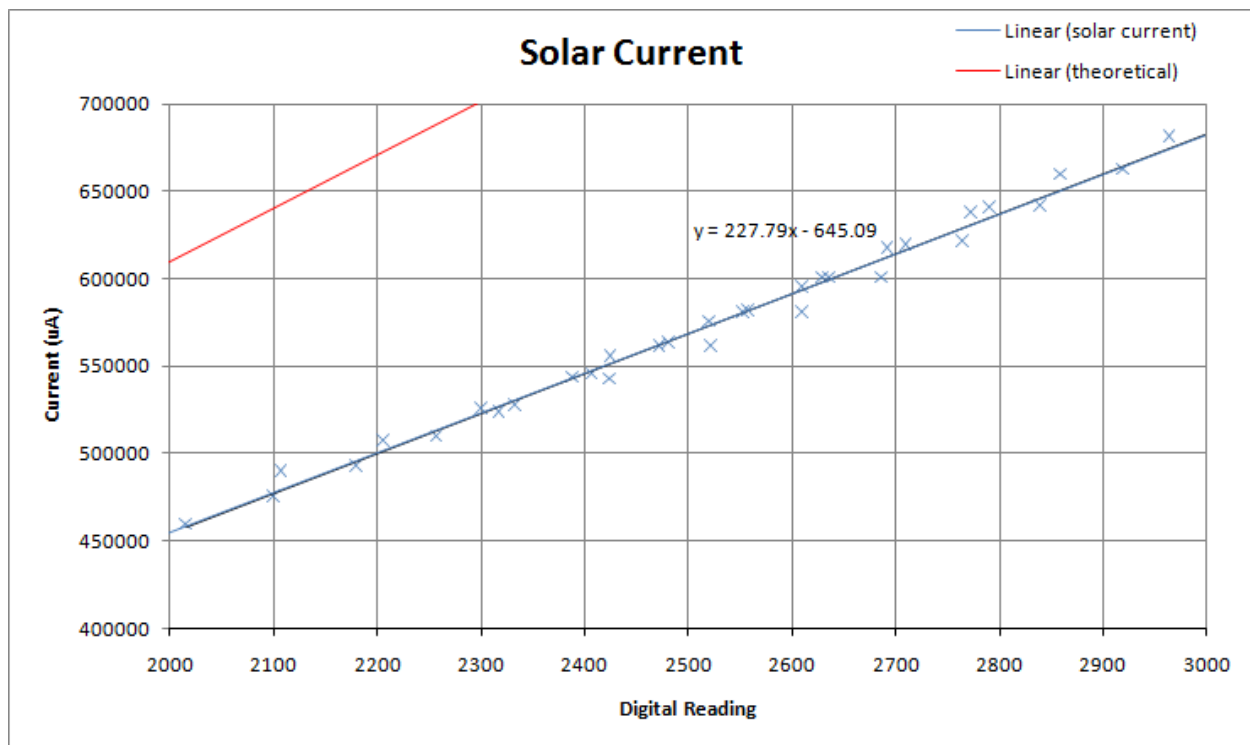
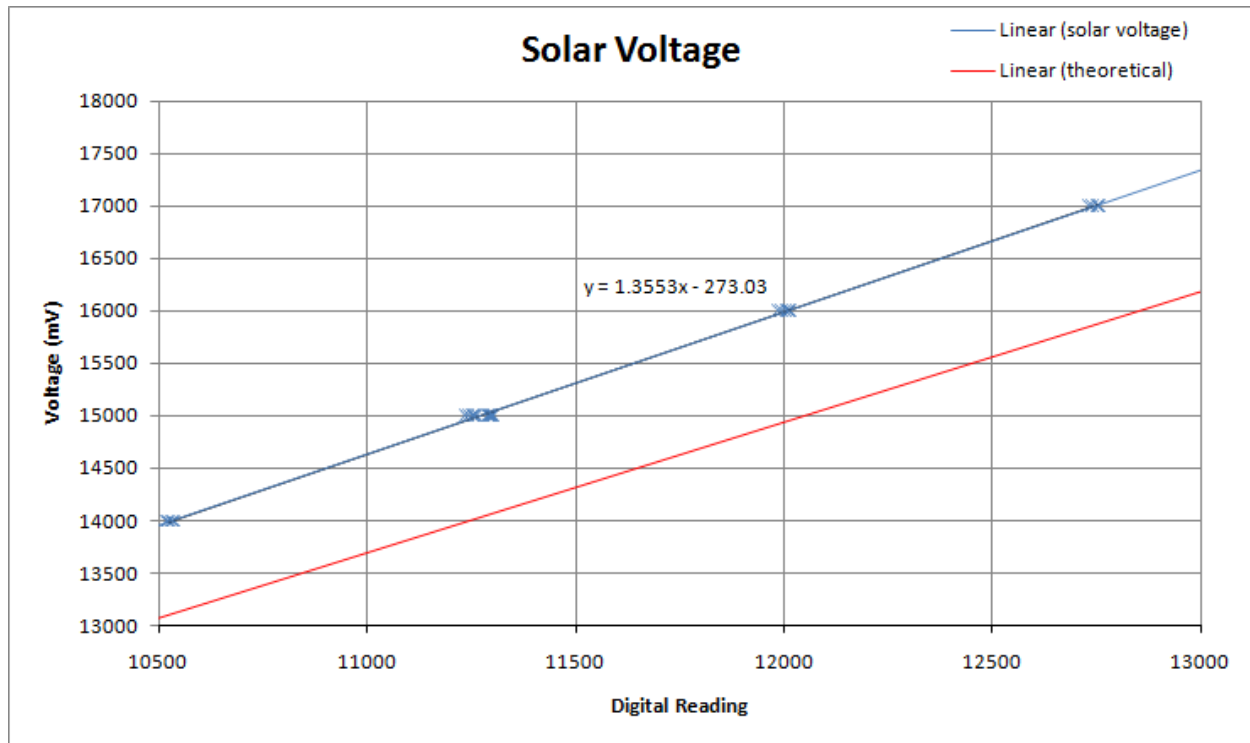
    if(delay >= totaldelay) {

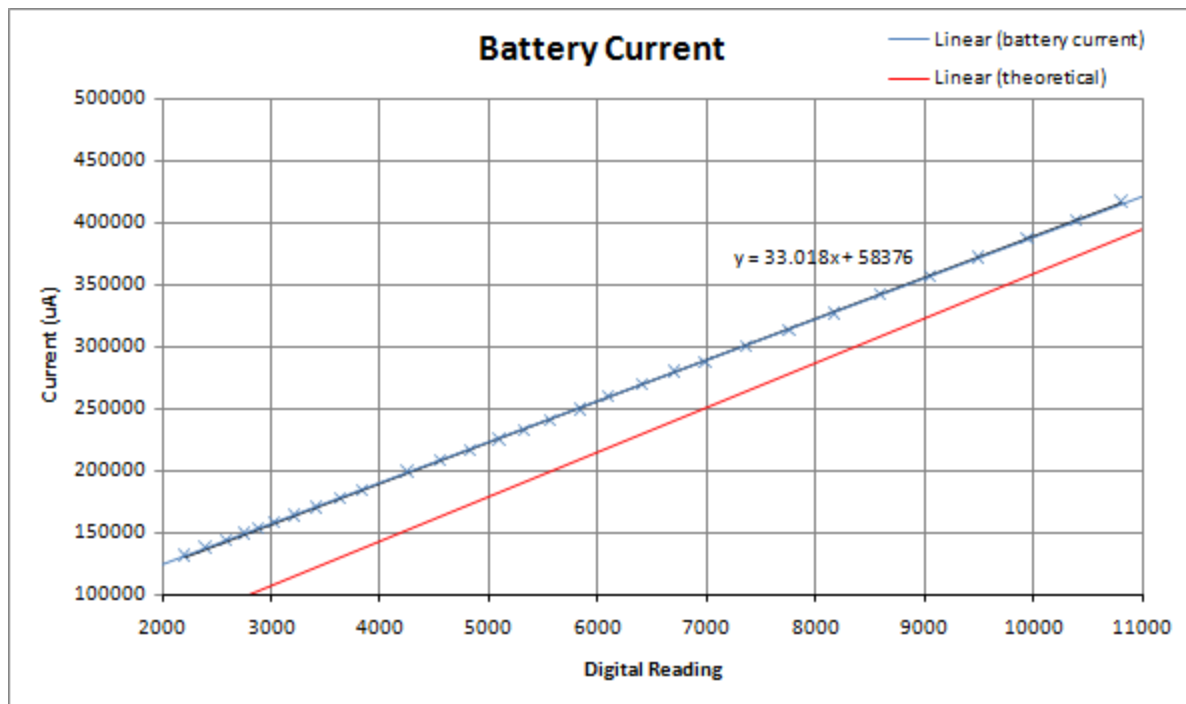
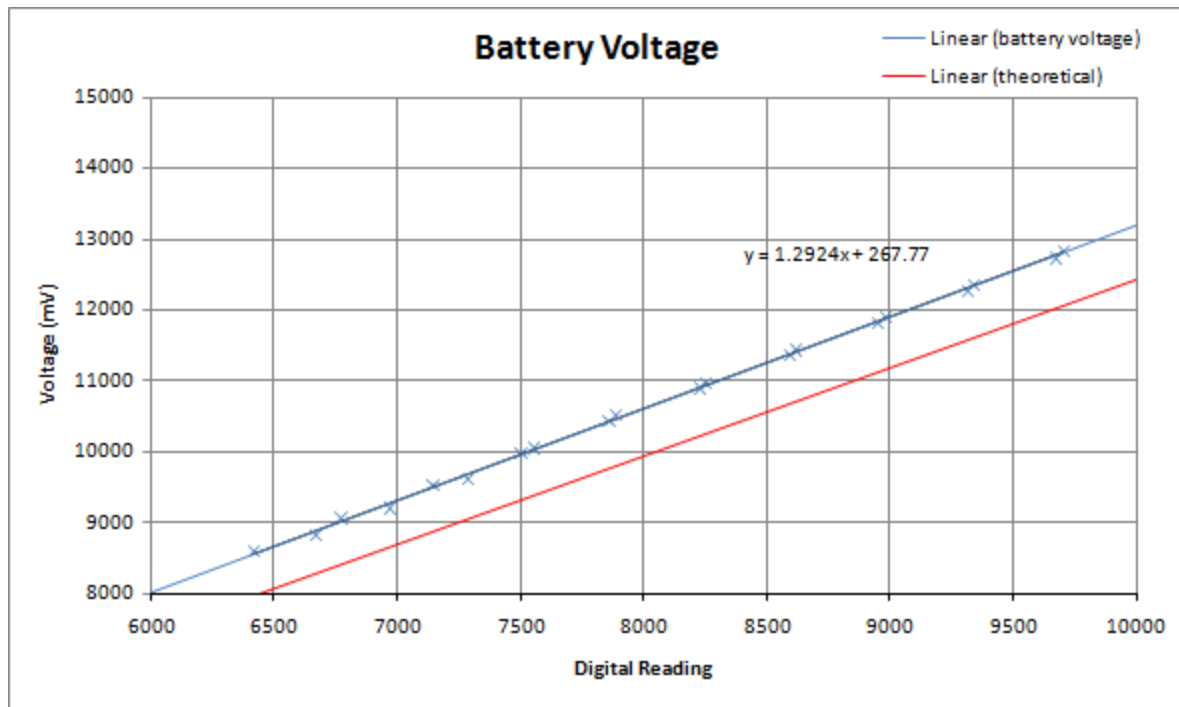
        //Used to verify Baudrate
        //P1OUT ^= 0x01;                // Toggle P1.0 using exclusive-OR

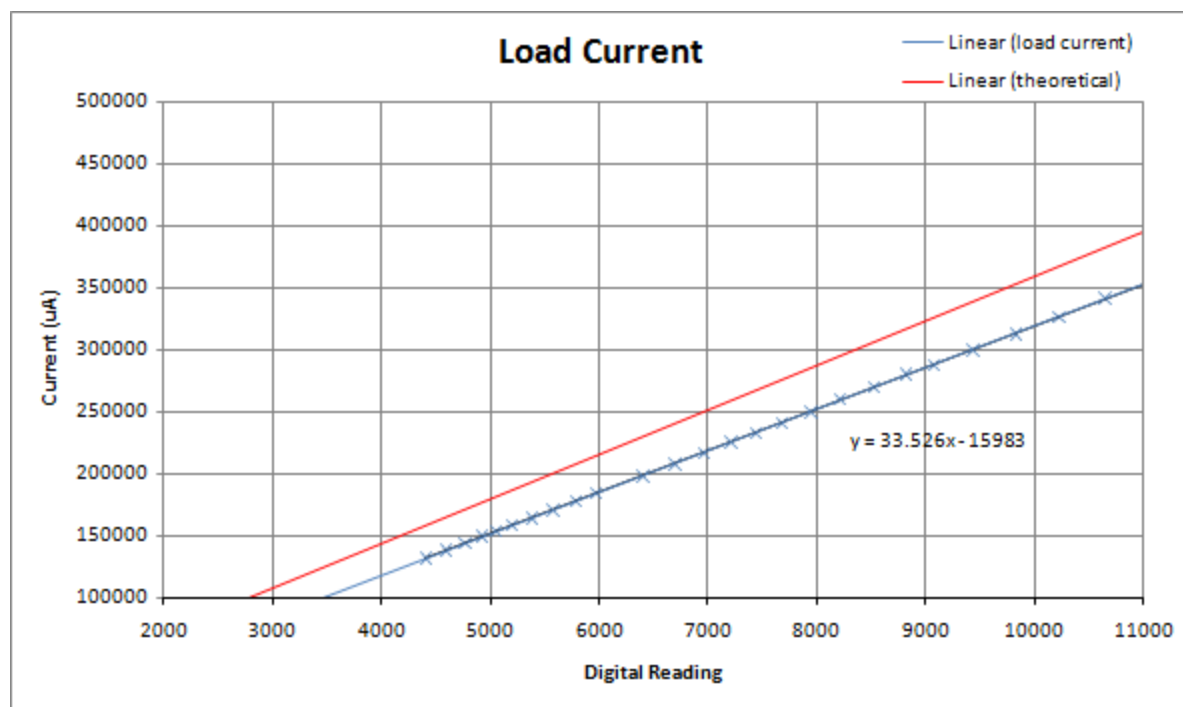
        switch(usart_state)
        {
            case 0: //do nothing, data not available
                break;
            case 1: //send start bit
                P2OUT &= 0xBF;
                usart_state += usart_state;
                break;
            case 2: //transmit data
                if(data & 0x01) P2OUT |= 0x40;
                else P2OUT &= 0xBF;
                data = data >> 1;
                i += 1;
                if(i >= 8)
                {
                    i = 0;
                    usart_state += 2;
                }
                break;
            case 4: //send stop bits
                P2OUT |= 0x40;
                i += 1;
                if(i >= 7)
                {
                    i = 0;
                    usart_state = 0;
                }
                break;
        }
        delay=0; //reset delay
    }
    delay+=1; //increment delay
}

```

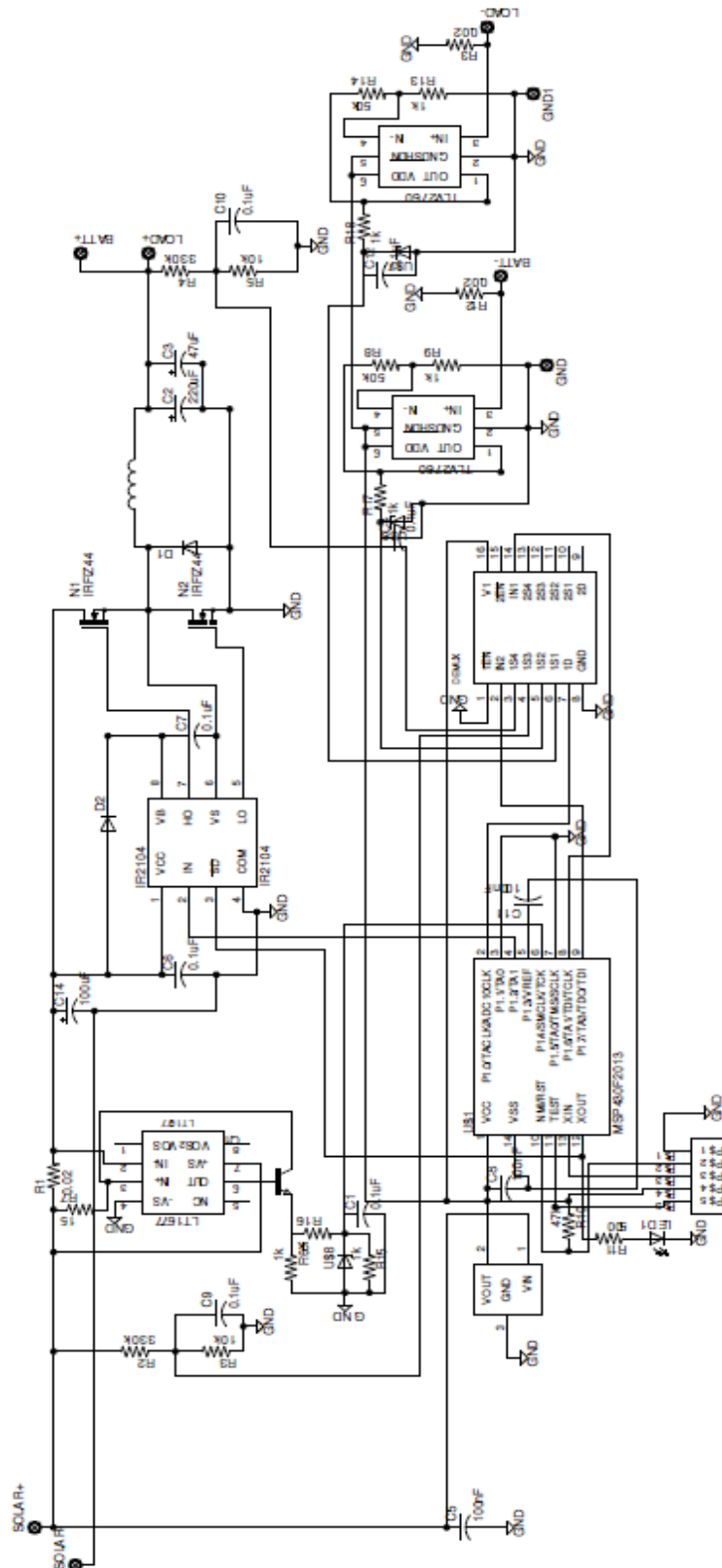
Appendix II – PCB Sensor Calibration Graphs

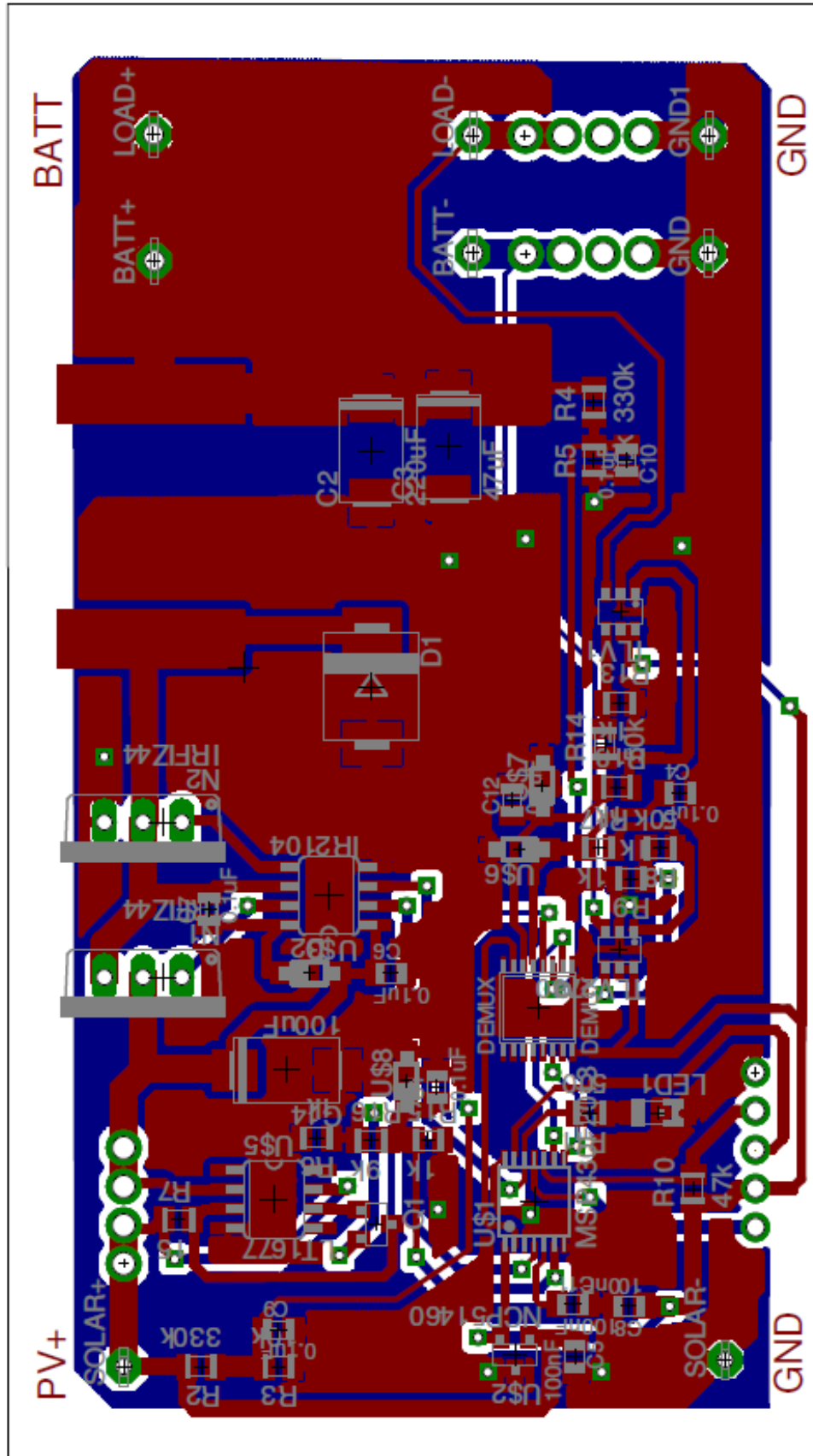


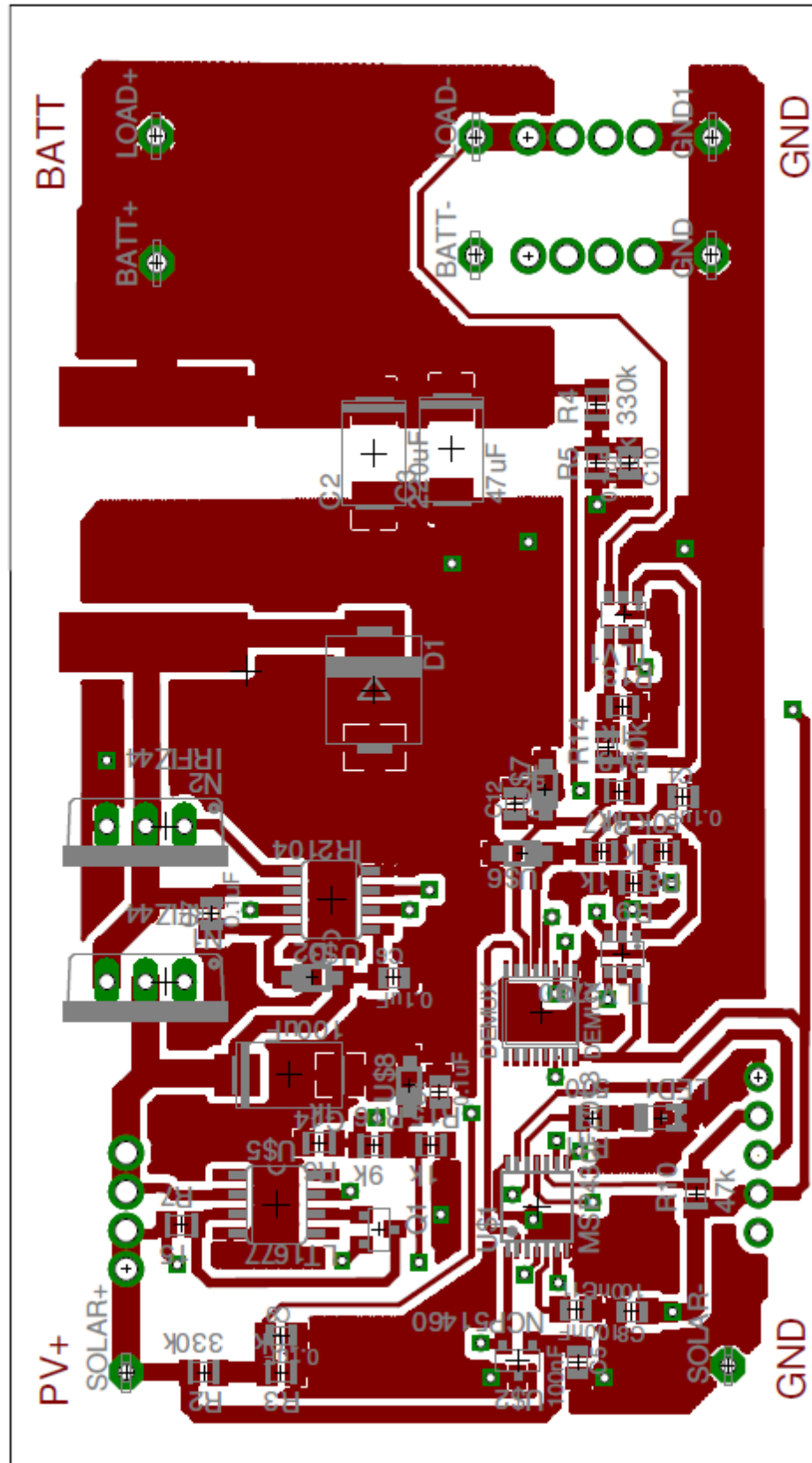


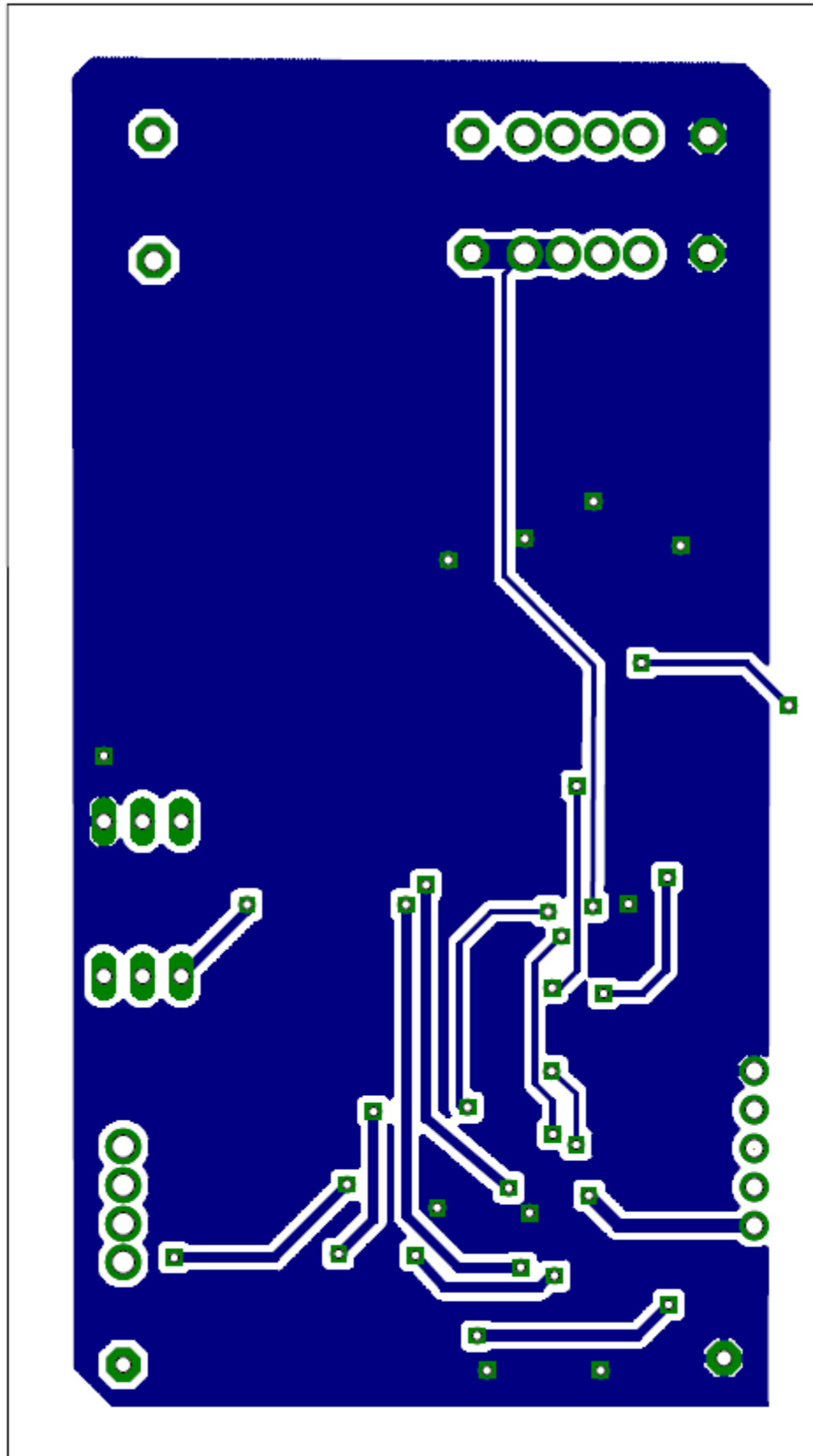


Appendix III – Schematic and PCB Layout









Appendix IV – User Manual

492 Project title: Mobile Power Station

492 Project team: Brad Jensen, Will Klema, Nate Schares, Dr. Ayman Fayed

Authors of the document: Kaila Krieser, Craig Christofferson, Mark Wisted, Dan Montgomery

1. What is the high level objective of the project?

The Mobile Power Station (MPS) is a device that manages and stores solar energy in a lithium ion (LI-ion) battery pack while tracking the maximum power point (MPPT) of the connected solar array. The MPS utilizes a DC buck converter design to control the impedance of the Li-ion pack and external load, allowing real-time control of the PWM signal and output voltage. Our client, PowerFilm Inc., expressed a need for the MPS in military and consumer markets.

2. What are the key functional requirements of the system?

- a. Solar panel input – 1.2A @ 5.4V (20W) amorphous silicon panel
- b. 100W minimum Li-ion battery capacity
- c. 15V DC input (with AC/DC Adapter)
- d. 12V DC output
- e. Operation in temperature range of -20° and 60° C
- f. Charge balancing circuitry to keep Li-ion batteries balanced
- g. Achieve 80% or greater efficiency
- h. Weight less than 5 pounds
- i. Manufacturing cost of under \$500
- j. Easily fit inside military ruck

3. What has been actually implemented?

The basis of the MPS is a buck converter, which controls the voltage to the battery pack, and thus the current flowing to pack. The buck converter is controlled by varying the duty cycle of a PWM signal generated and controlled by the MSP430. Programmed in C code, the MSP430 follows a charging algorithm based on feedback voltages and currents. UART communication to a PC provides diagnostic information and the charge controller balances the Li-ion battery pack.

4. How to setup the system?

- a. Connect serial communication
- b. Connect source (solar panel for use or DC source for testing purposes)
- c. Enable graphical user interface (GUI)
- d. Connect battery pack or other external load
- e. Monitor voltage and current fluctuation at the source and load

5. Test results observed?

Li-ion battery charging cycle includes three states: trickle, constant current and constant voltage. The charging cycle must be strictly followed to prevent overcharge and fire. Measuring the output voltage and current of the buck converter, the MPS changes the duty cycle of the buck converter to follow the charging cycle. At the same time, the MPS measures the solar input voltage and current to calculate the MPP. If too much current is demanded from the solar array (during constant current charging), the MPS will decrement the duty cycle to shift the MPP, maintaining maximum area under the solar I-V curve.

The demonstration showed accurate and safe, real-time function of the MPS, as displayed on the accompanying GUI. Traces of MPS voltage, current, and PWM waveforms were displayed and easily read.

6. Critique of the project.

- a. Provide at least one strength and at least one weakness of the implemented system
 - i. Strength
 1. The strengths and dynamics of the group help the success of the project. Each one of the group members was an expert in an area of the project.
 2. The team is concerned with safety; therefore, multiple tests are being conducted to ensure that the lithium ion batteries will not overcharge.
 3. Each member of the team is knowledgeable of the various aspects of the project. The questions asked were answered thoroughly.
 4. It appeared that the group completed detailed plans that enabled them to minimize the roadblocks encountered.
 - ii. Weaknesses
 1. The end-product could have been refined further if the group realized the breadboard was not working properly sooner in the project.
- b. Does the implementation meet the specification?
 - i. Yes, the implementation meets the specifications. The design is nearly 86% efficient and the team will be competing in a Texas Instruments competition.